

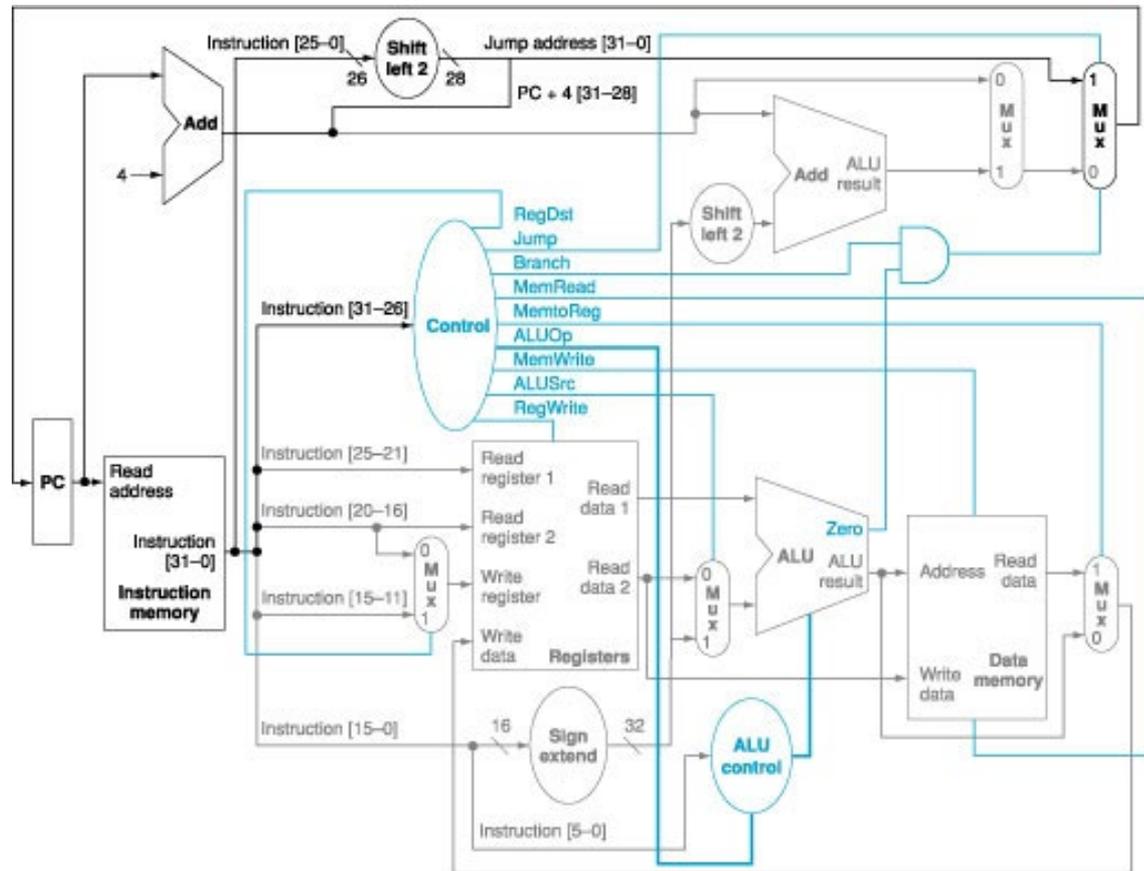
# Lecture 5 - Modeling for Synthesis

## Register Transfer Level (RTL) Design

# Register Transfer Language (RTL) Design

- A **system** is viewed as a structure comprising registers, functions and their control signals
- Show dataflow through the system
- Instructions, Data, Addresses
- Functions store and manipulate data

No gates!!!



# RTL register model

---

- Model register to hold one datum of some type
- Individual bits are not manipulated

```
library ieee; use ieee.std_logic_1164.all;
```

```
entity Reg8 is
```

```
  port (D: in std_logic_vector(7 downto 0);  
        Q: out std_logic_vector(7 downto 0);  
        LD: in std_logic);
```

```
end Reg8;
```

```
architecture behave of Reg8 is
```

```
begin
```

```
  process(LD)
```

```
  begin
```

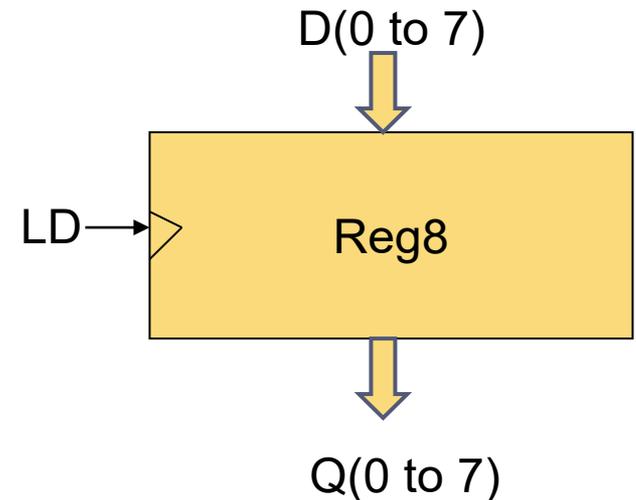
```
    if (LD'event and LD='1') then
```

```
      Q <= D; -- load data into the register
```

```
    end if;
```

```
  end process;
```

```
end;
```



# Asynchronous control inputs

```
library ieee; use ieee.std_logic_1164.all;
```

```
entity Reg8 is
```

```
  port (D: in std_logic_vector (7 downto 0);
```

```
        CLK,PRE,CLR: in bit;
```

```
        Q: out std_logic_vector (7 downto 0));
```

```
end Reg8;
```

```
architecture behave of Reg8 is
```

```
begin
```

```
  process(clk, PRE, CLR)
```

```
  begin
```

```
    if (CLR='0') then
```

```
      Q <="00000000";
```

```
    elsif (PRE='0') then
```

```
      Q <= (others => '1');
```

```
    elsif rising_edge (clk) then
```

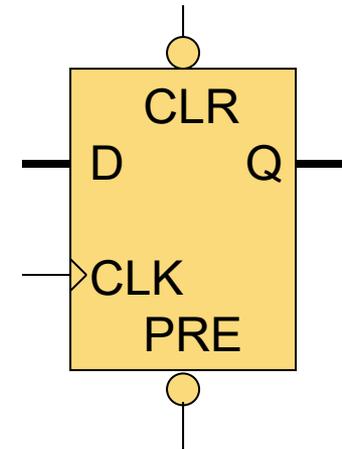
```
      Q <= D;
```

```
    end if;
```

```
  end process;
```

```
end;
```

--Async PRE/CLR



-- async CLR has precedence

-- force register to all 0s

-- async PRE has precedence if CLR='0'

-- force register to all 1s

-- sync operation only if CLR=PRE='1'

-- load D on clock transition

# Synchronous reset/set

---

--Reset function triggered by clock edge

```
process (clk)
begin
    if (clk'event and clk = '1') then
        if reset = '1' then
            Q <= "00000000" ;
        else
            Q <= D ;
        end if;
    end if;
end process;
```

# Register with enable

---

```
process (clk)
begin
    if rising_edge(clk) then -- detect clock transition
        if enable = '1' then -- enable load on clock transition
            Q <= D ;
        end if;
    end if;
end process;
```

# Register with parameterized width

---

-- One model of a given function with variable data size

```
library ieee; use ieee.std_logic_1164.all;
```

```
entity REGN is
```

```
    generic (N: integer := 8);                -- N specified when REG used
```

```
    port ( CLK, RST, PRE, CEN: in std_logic;
```

```
          DATAIN: in std_logic_vector (N-1 downto 0); -- N-bit data in
```

```
          DOUT: out std_logic_vector (N-1 downto 0) -- N-bit data out
```

```
    );
```

```
end entity REGN;
```

```
architecture RTL of REGN is
```

```
begin
```

```
    process (CLK) begin
```

```
        if (CLK'event and CLK = '1') then
```

```
            if (RST = '1') then DOUT <= (others => '0'); --reset to all 0s
```

```
            elsif (PRE = '1') then DOUT <= (others => '1'); --preset to all 1s
```

```
            elsif (CEN = '1') then DOUT <= DATAIN; --load data
```

```
            end if;
```

```
        end if;
```

```
    end process;
```

```
end architecture RTL;
```

Vectors: "100" = ('1','0','0') = ('1', others => '0')

Arbitrarily long: "00...0" = (others => '0')

# Instantiating the parameterized register

---

```
library ieee; use ieee.std_logic_1164.all;
```

```
entity TOP is
```

```
    port (    CLK,X,Y,A,B,C: in std_logic;
            DIN: in  std_logic_vector(5 downto 0);
            Q1: out std_logic_vector(5 downto 0);
            Q2: out std_logic_vector(4 downto 0);
            Q3: out std_logic_vector(3 downto 0)
    );
```

```
end entity TOP;
```

```
architecture HIER of TOP is
```

```
component REGN is
```

```
    generic (N: integer := 8);
    port (    CLK, RST, PRE, CEN: in std_logic;
            DATAIN: in  std_logic_vector (N-1 downto 0);
            DOUT:   out std_logic_vector (N-1 downto 0)
    );
```

```
end component REGN;
```

```
begin
```

```
R1: REGN generic map (6) port map  --6-bit register
      (CLK, A, B, C, DIN, Q1);
```

```
R2: REGN generic map (5) port map  --5-bit register (low 5 bits of DIN)
      (CLK, Y, X, C, DIN(4 downto 0),Q2);
```

```
R3: REGN generic map (4) port map  --4-bit register (low 4 bits of DIN)
      (CLK=>CLK, RST=>A, PRE=>B, CEN=>C, DATAIN=>DIN(3 downto 0), DOUT=>Q3);
```

```
end architecture HIER;
```

## 2-to-1 mux with parameterized data size

---

entity muxN is

```
generic (N: integer := 32); -- data size parameter
port ( A,B: in  std_logic_vector(N-1 downto 0);
      Y:  out std_logic_vector(N-1 downto 0);
      Sel: in  std_logic);
```

end muxN;

architecture rtl of muxN is

begin

```
Y <= A when Sel = '0' else B; -- A,B,Y same type
```

end;

-- specify parameter N at instantiation time

M: muxN generic map (16)

```
port map(A=>In1, B=>In2, Y=>Out1);
```

# Other types of generic parameters

---

entity and02 is

```
    generic (Tp : time := 5 ns);    -- gate delay
```

parameter

```
    port (A,B: in std_logic;
          Y:  out std_logic);
```

end and02;

architecture eqn of and02 is

begin

```
    Y <= A and B after Tp;    -- gate with delay Tp
```

end;

.....

```
A_tech1: and02 generic map (2 ns) port map (M,N,P);
```

```
A_tech2: and02 generic map (1 ns) port map (H,K,L);
```

← Gates with  
different delays.

# IEEE Std. 1076.3 Synthesis Libraries

---

- **Supports arithmetic models**
  - `ieee.numeric_std` (ieee library package)
    - defines UNSIGNED and SIGNED types as arrays of `std_logic`

```
type SIGNED is array(NATURAL range <>) of STD_LOGIC;  
type UNSIGNED is array(NATURAL range <>) of STD_LOGIC;
```
    - defines arithmetic/relational operators on these types
  - Supports RTL models of functions
  
- **Lesser-used packages:**
  - `ieee.numeric_bit`
    - same as above except SIGNED/UNSIGNED are arrays of type `bit`
  - `ieee.std_logic_arith` (*from Synopsis*)
    - Non-standard predecessor of `numeric_std/numeric_bit`

# NUMERIC\_STD package contents

---

- Arithmetic functions: + - \* / rem mod
  - Combinations of operand types for which operators are defined:
    - SIGNED + SIGNED return SIGNED
    - SIGNED + INTEGER return SIGNED
    - INTEGER + SIGNED return SIGNED
    - SIGNED + STD\_LOGIC return SIGNED
  - PLUS: above combinations with UNSIGNED and NATURAL
- Other operators for SIGNED/UNSIGNED types:
  - Relational: = /= < > <= >=
  - Shift/rotate: sll, srl, sla, sra, rol, ror
  - Maximum(a,b), Minimum(a,b)
- Convert between types:
  - TO\_INTEGER(SIGNED), TO\_INTEGER(UNSIGNED)
  - TO\_SIGNED(INTEGER,#bits), TO\_UNSIGNED(NATURAL,#bits)
  - RESIZE(SIGNED or UNSIGNED,#bits) – changes # bits in the vector

# Arithmetic with NUMERIC\_STD package

---

```
library IEEE;
use IEEE.STD_LOGIC_1164.all;
use IEEE.NUMERIC_STD.all;
entity Adder4 is
    port ( in1, in2 : in  UNSIGNED(3 downto 0) ;
          mySum : out UNSIGNED(3 downto 0) ) ;
end Adder4;

architecture Behave_B of Adder4 is
begin
    mySum <= in1 + in2; -- overloaded '+' operator
end Behave_B;
```

UNSIGNED = UNSIGNED + UNSIGNED

# Conversion of “closely-related” types

---

- **STD\_LOGIC\_VECTOR, SIGNED, UNSIGNED:**
  - All arrays of **STD\_LOGIC** elements
  - Example: How would one interpret “1001” ?
    - **STD\_LOGIC\_VECTOR:** simple pattern of four bits
    - **SIGNED:** 4-bit representation of number -7 (**2’s complement #**)
    - **UNSIGNED:** 4-bit representation of number 9 (**unsigned #**)
- **Vectors of same element types can be “converted” (re-typed/re-cast) from one type to another**
  - signal A: `std_logic_vector(3 downto 0) := “1001”;`
  - signal B: `signed(3 downto 0);`
  - signal C: `unsigned(3 downto 0);`
  - `B <= signed(A);`                    -- interpret A value “1001” as number -7
  - `C <= unsigned(A);`                -- interpret A value “1001” as number 9
  - `A <= std_logic_vector(B);`        -- interpret B as bit pattern “1001”

# Conversion of “closely-related” types

For arrays of same dimension, *having elements of same type*

library IEEE;

use IEEE.STD\_LOGIC\_1164.all;

use IEEE.NUMERIC\_STD.all;

entity Adder4 is

```
port ( in1, in2 : in  STD_LOGIC_VECTOR(3 downto 0) ;  
      mySum : out STD_LOGIC_VECTOR(3 downto 0) ) ;
```

end Adder4;

architecture Behave\_B of Adder4 is

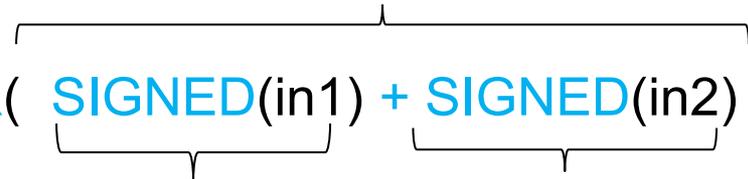
begin

```
mySum <=
```

```
  STD_LOGIC_VECTOR( SIGNED(in1) + SIGNED(in2) );
```

```
end Behave_B;
```

**SIGNED** result



Interpret **STD\_LOGIC\_VECTOR** as **SIGNED**  
Function: **SIGNED** = **SIGNED** + **SIGNED**

Interpret **SIGNED** result as **STD\_LOGIC\_VECTOR**.

# Example – binary counter

---

```
library IEEE;
use IEEE.STD_LOGIC_1164.all;
use IEEE.NUMERIC_STD.all;
ENTITY counter IS
    port( Q: out std_logic_vector(3 downto 0);
        ....
END counter;
```

```
ARCHITECTURE behavior OF counter IS
    signal Qinternal: unsigned(3 downto 0);
begin
```

From NUMERIC\_STD package



```
    Qinternal <= Qinternal + 1;    -- UNSIGNED = UNSIGNED + NATURAL
    Q <= std_logic_vector(Qinternal); -- re-type unsigned as std_logic_vector
```

# Using a “variable” to describe sequential behavior within a process

---

-- Assume Din and Dout are std\_logic\_vector

-- and numeric\_std package is included

```
cnt: process(clk)
```

```
    variable count: integer; -- internal counter state
```

```
begin -- valid only within a process
```

```
    if clk='1' and clk'event then
```

```
        if ld='1' then
```

```
            count := to_integer(unsigned(Din)); --update immediately
```

```
        elsif cnt='1' then
```

```
            count := count + 1; --update immediately
```

```
        end if;
```

```
    end if;
```

```
    Dout <= std_logic_vector(to_unsigned(count,32)); --schedule Dout
```

```
end process;
```

# Counting to some max\_value (not $2^n$ )

---

-- full-sized comparator circuit generated to check count = max

**process begin**

**wait** until clk'event and clk='1' ;

**if** (count = max\_value) **then**

    count <= 0 ;                   --roll over from max\_value to 0

**else**

    count <= count + 1 ;   --otherwise increment

**end if** ;

**end process** ;

# Decrementer and comparator

---

**process begin**

**wait** until clk'event and clk='1' ;

**if** (count = 0) **then**

    count <= max\_value ; -- roll over from 0 to max\_value

**else**

    count <= count - 1 ; -- otherwise decrement

**end if** ;

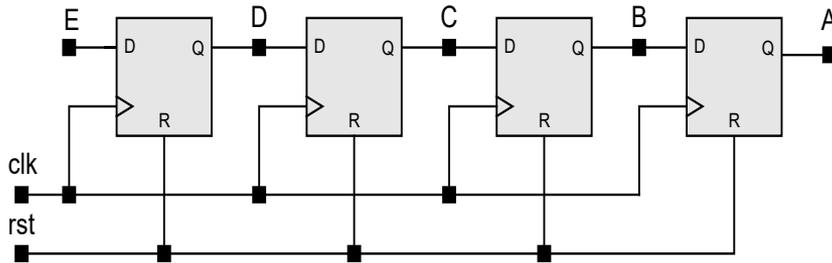
**end process** ;

# Verilog Modeling Trap

---

- The order of execution of **procedural statements** in a cyclic behavior may **depend on the order** in which the statements are listed
- Procedural assignments are called **“blocked” assignments** (or blocking assignments)
  - Execute sequentially
  - A procedural assignment must complete execution before the next statement can be executed
  - i.e. the statements that follow a procedural statement are **“blocked”** till the current one completes execution
- Expression substitution is recognized by synthesis tools

# Example: Modeling Trap of a Shift Register



```

module shiftreg_PA (E, A, clk, rst);
  output A;
  input E;
  input clk, rst;
  reg A, B, C, D;

  always @ (posedge clk or posedge rst)
  begin
    if (rst) begin
      A = 0; B = 0; C = 0; D = 0; end
    else begin
      A = B;
      B = C;
      C = D;
      D = E;
    end
  end
endmodule
  
```

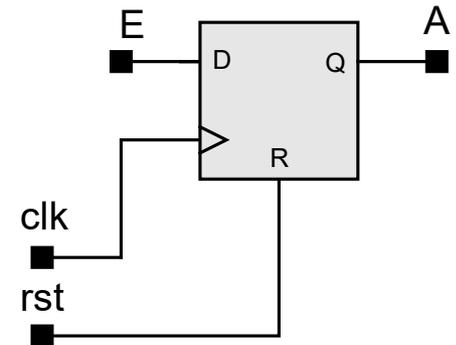
```

module shiftreg_PA_rev (E, A, clk, rst);
  output A;
  input E;
  input clk, rst;
  reg A, B, C, D;

  always @ (posedge clk or posedge rst)
  begin
    if (rst) begin
      A = 0; B = 0; C = 0; D = 0; end
    else begin
      D = E;
      C = D;
      B = C;
      A = B;
    end
  end
endmodule
  
```

D = E;  
 C = D;  
 B = C;  
 A = B;

} **A=E**



# Nonblocking Assignment ( $\leq$ ) in Cyclic Behavior

---

- Effectively execute concurrently rather than sequentially by blocked assignments
  - Independent of the order where they are listed
- **Simulator must**
  - **Sample all variables referenced by RHS with nonblocking assignments**
  - **Held them in memory**
  - **Use them to update LHS variables **concurrently****
    - **Before the assignments are evaluated**
  - **Nonblocking makes NO dependency between statements**
- Avoid having multiple behaviors assigning values to be the same variable
  - Otherwise, software race condition makes outcome indeterminate
  - For example, multi-driver case

# Blocked (=) v.s. Nonblocking (<=)

---

- If no data dependency, results of blocked and nonblocking assignments are identical
- Strongly recommend
  - Blocked assignment for combinational logic using level sensitive behavior
  - Nonblocking assignments for edge sensitive behavior

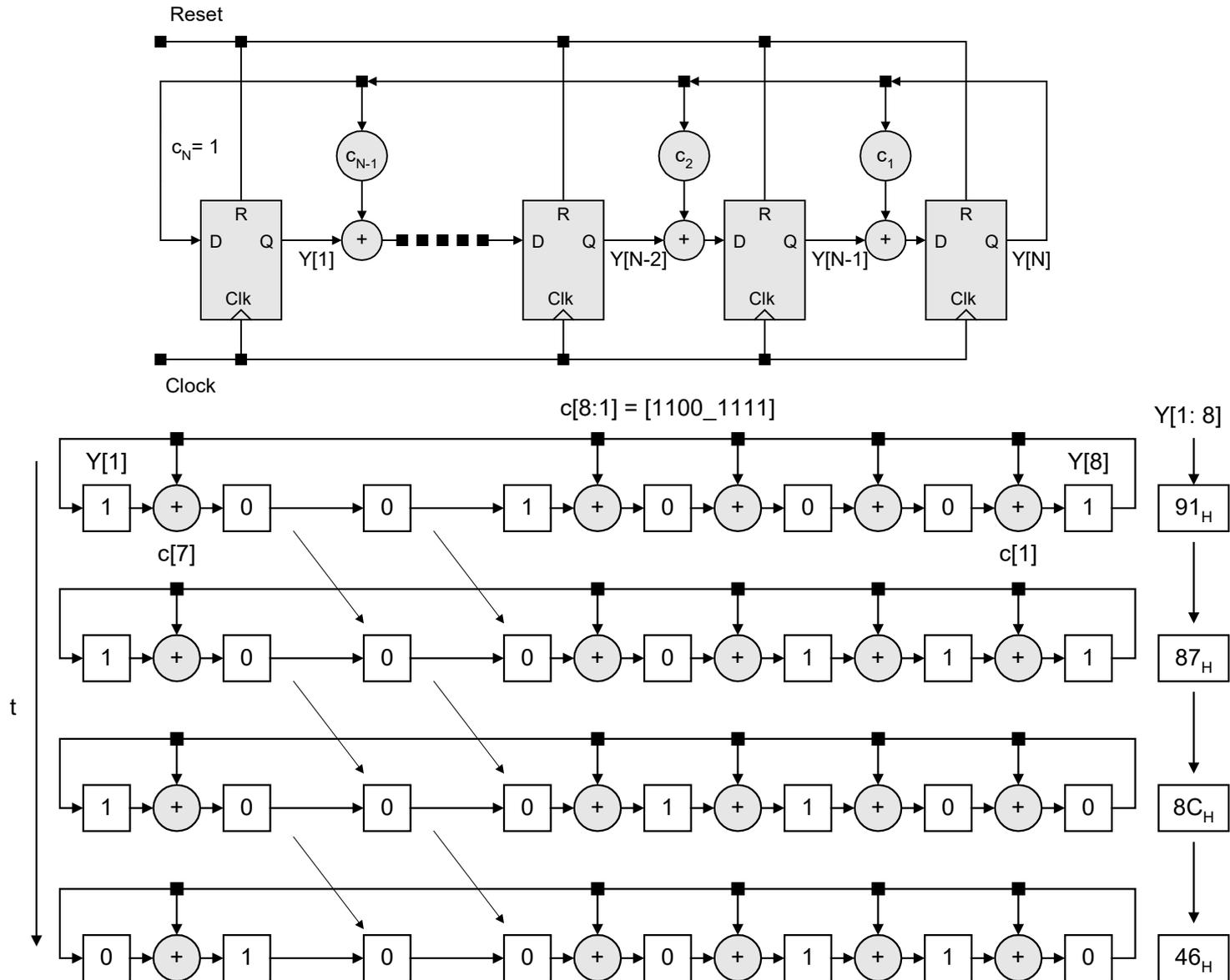
# Shift Register Using Nonblocking Assignments

---

```
module shiftreg_nb (A, E, clk, rst);
    output A;
    input  E;
    input  clk, rst;
    reg    A, B, C, D;

    always @ (posedge clk or posedge rst)
    begin
        if (rst)
            begin A <= 0; B <= 0; C <= 0; D <= 0;
            end
        else
            begin
                A <= B;           //      D <= E;
                B <= C;           //      C <= D;
                C <= D;           //      B <= C;
                D <= E;           //      A <= B;
            end
        end
    end
endmodule
```

# Linear-Feedback Shift Register (Type II LFSR) Dataflow



# LFSR --- RTL Dataflow

---

```
module Auto_LFSR_RTL (Y, Clock, Reset);
  parameter                               Length = 8;
  parameter [1: Length]                   initial_state = 8'b1001_0001; // 91h
  parameter [1: Length]                   Tap_Coefficient = 8'b1111_0011;

  input                                    Clock, Reset;
  output [1: Length]                       Y;
  reg [1: Length]                           Y;

  always @ (posedge Clock)
    if (!Reset) Y <= initial_state;           // Active-low reset to initial state
    else begin
      Y[1] <= Y[8];
      Y[2] <= Tap_Coefficient[7] ? Y[1] ^ Y[8] : Y[1];
      Y[3] <= Tap_Coefficient[6] ? Y[2] ^ Y[8] : Y[2];
      Y[4] <= Tap_Coefficient[5] ? Y[3] ^ Y[8] : Y[3];
      Y[5] <= Tap_Coefficient[4] ? Y[4] ^ Y[8] : Y[4];
      Y[6] <= Tap_Coefficient[3] ? Y[5] ^ Y[8] : Y[5];
      Y[7] <= Tap_Coefficient[2] ? Y[6] ^ Y[8] : Y[6];
      Y[8] <= Tap_Coefficient[1] ? Y[7] ^ Y[8] : Y[7];
    end
endmodule
```

# LFSR --- RTL Repetitive Algorithm

---

```
module Auto_LFSR_ALGO (Y, Clock, Reset);
  parameter                               Length = 8;
  parameter [1: Length]                   initial_state = 8'b1001_0001;
  parameter [1: Length]                   Tap_Coefficient = 8'b1111_0011;
  input                                     Clock, Reset;
  output [1: Length]                       Y;
  integer                                   Cell_ptr;
  reg                                       Y;

  always @ (posedge Clock)
  begin
    if (Reset == 0) Y <= initial_state;    // Arbitrary initial state, 91h
    else
      begin
        for (Cell_ptr = 2; Cell_ptr <= Length; Cell_ptr = Cell_ptr + 1)
          if (Tap_Coefficient [Length - Cell_ptr + 1] == 1)
            Y[Cell_ptr] <= Y[Cell_ptr - 1]^ Y [Length]; // ^ is xor
          else
            Y[Cell_ptr] <= Y[Cell_ptr - 1];
        Y[1] <= Y[Length];
      end
    end
endmodule
```

# Verilog Repetitive Statements

---

- for, repeat, while, forever
  - All activities of all iterations are done in one time step
  - “disable” to terminate a named block
  - Some logic synthesis tools can only synthesize “for” loop
    - i.e., repeat, while, forever are not synthesizable in these tools

# Verilog Statement

---

- Statement can be
  - a single statement or
  - a block statement

```
begin  
    statement1  
    statement2  
    ...  
end
```

- A named block statement

```
begin: <block_name>  
    statement1  
    statement2  
    ...  
end
```

# Ones Counter

---

- Verilog bitwise right-shift operator ( $\gg$ ), filling with '0'
  - Arithmetic right-shift ( $\ggg$ )
- Compare the following two designs

*// count\_of\_1s* declares a named block of statements

*// Original design*

**begin:** *count\_of\_1s*

**reg** [7: 0] temp\_reg;

count = 0;

temp\_reg = reg\_a; // load a data word

**while** (temp\_reg)

**begin**

**if** (temp\_reg[0])

    count = count + 1;

    temp\_reg = temp\_reg  $\gg$  1;

**end**

**end**

*// Alternative*

**begin:** *count\_of\_1s*

**reg** [7: 0] temp\_reg;

count = 0;

temp\_reg = reg\_a; // load a data word

**while** (temp\_reg)

**begin**

    count = count + temp\_reg[0];

    temp\_reg = temp\_reg  $\gg$  1;

**end**

**end**

# Find\_First\_One

---

- Find the location of the first 1 in a 16-bit word
  - The word is assumed to contain at least one 1

```
module find_first_one (index_value, A_word, trigger);
  output      [3: 0]    index_value;
  input       [15: 0]   A_word;
  input                               trigger;
  reg         [3: 0]    index_value;

  always @ (trigger)
    begin: search_for_1
      index_value = 0;
      for (index_value = 0; index_value <= 15; index_value = index_value + 1)
        if (A_word[index_value] == 1)
          disable search_for_1;
    end
endmodule
```

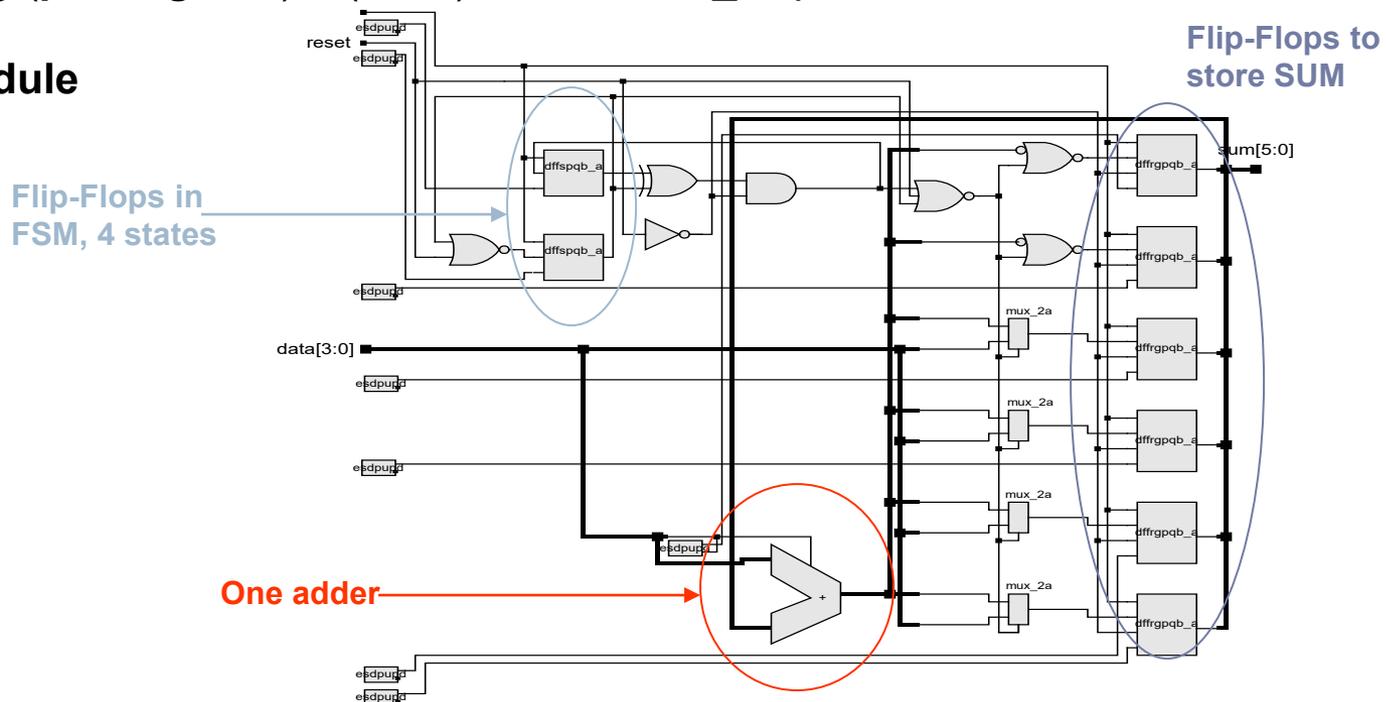
# Multicycle Operations -- 4-cycle Adder

---

- Some digital machines have repetitive operations distributed over multiple clock cycles
  - Can be modeled in Verilog by a synchronous cyclic behavior that has as many nested edge-sensitive event control expressions as needed to complete the operations
  - May not be synthesizable
- Example: 4-cycle adder
  - To form the sum of four successive samples of a datapath
    - Store the samples in registers then use multiple adders
    - Or, one adder to accumulate the sum sequentially
      - One FSM to control the 4-cycle operation and only one adder
      - The resulting synthesized implementation
  - To ensure proper re-initialization, “disable” is in each clock cycle
    - Regardless when the “reset” is asserted

# 4-cycle Adder

```
module add_4cycle (sum, data, clk, reset);  
  output      [5: 0]  sum;  
  input       [3: 0]  data;  
  input       clk, reset;  
  reg         sum;  
  always @ (posedge clk) begin: add_loop  
    if (reset) disable add_loop; else sum <= data;  
    @ (posedge clk) if (reset) disable add_loop; else sum <= sum + data;  
    @ (posedge clk) if (reset) disable add_loop; else sum <= sum + data;  
    @ (posedge clk) if (reset) disable add_loop; else sum <= sum + data;  
  end  
endmodule
```



# Algorithmic State Machine (ASM) Charts

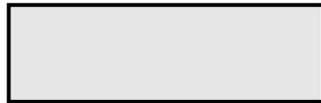
---

- State Transition Graphs (STGs)
  - Indicate the transitions that result from inputs applied to the state machine in a particular state
  - Do not directly display the evolution of states under the application of input data
- ASM Charts
  - Abstraction of functionality of a sequential machine
    - Reveal the sequential steps of a machine's activity
  - Focus on activities rather than content of storage elements
    - Example: the counter to be introduced shortly
      - Three states: idle, incrementing and decrementing
      - Independent of counter word width
  - ASM chart elements
    - **state box**
    - **decision box**
    - **conditional box**
  - Clock governs transitions between states
  - Linked ASM charts describe complex machines
    - ASM charts represent both Mealy and Moore machines

# ASM Chart Elements

---

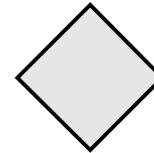
- State box
  - Each state box represents the state of the machine between synchronizing clock events
- Decision box
- Conditional box



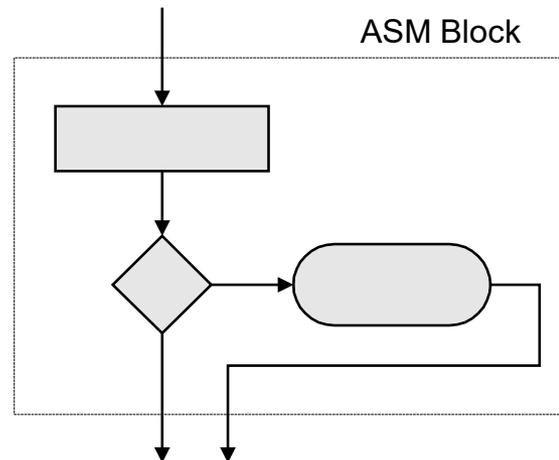
State Box



Conditional Output or  
Register Operation Box

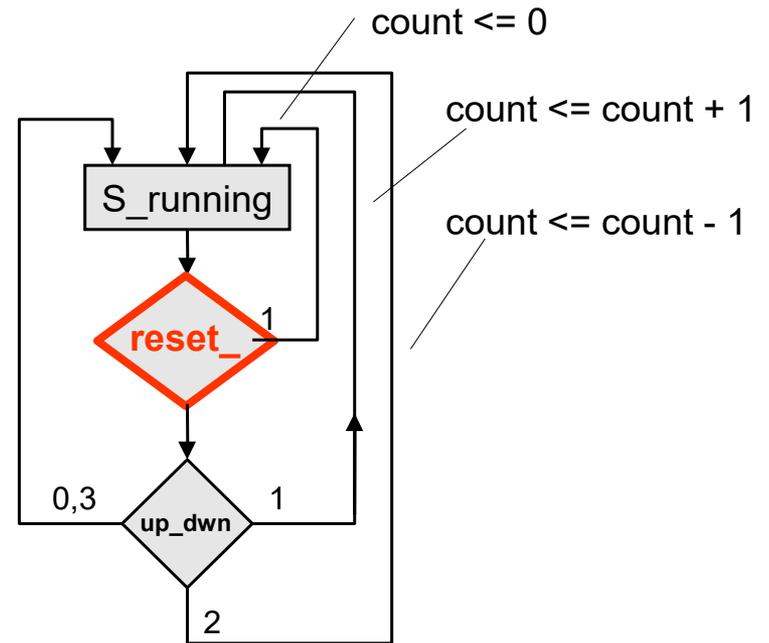
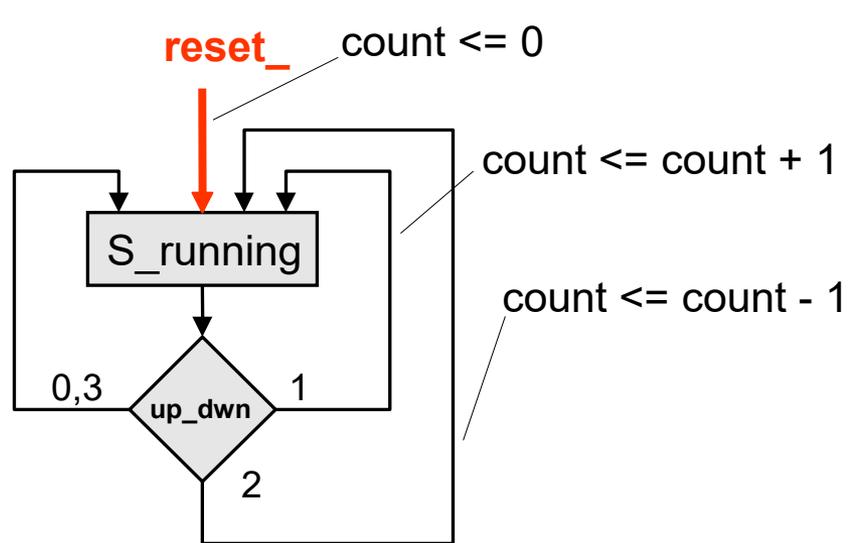


Decision Box



# Asyn/Synchronous Reset in ASM

- Asynchronous reset: a RESET input to the reset state box
- Synchronous reset: one decision box of RESET input



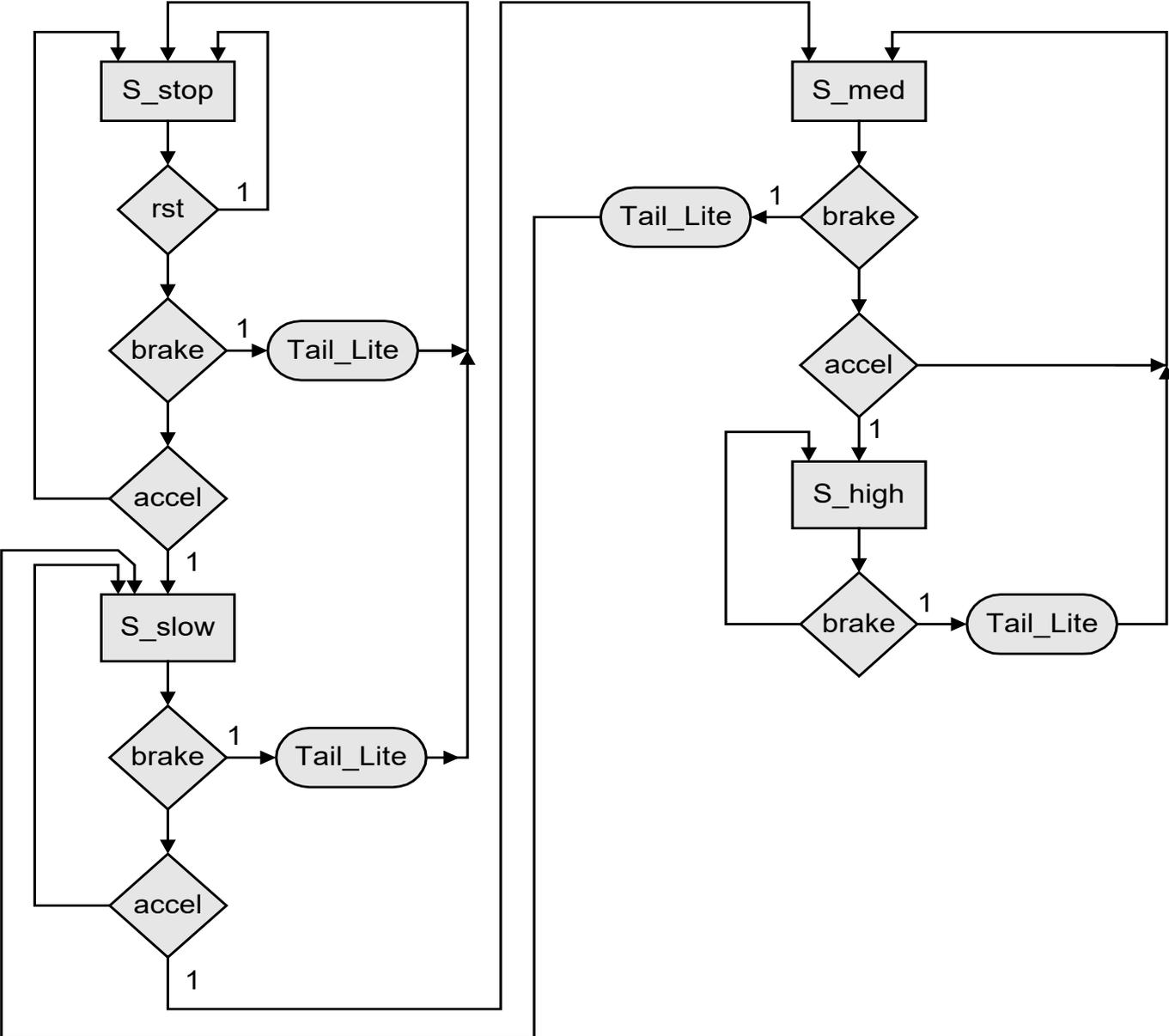
# ASM Chart (cont.)

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- Only paths leading to a change in states are shown in ASM
- If a variable not appear in a decision box on a path leaving a state, then the path is independent of the value of the variable

# ASM Chart Example: Tail Light Controller

## A Mealy Machine with Synchronous Reset



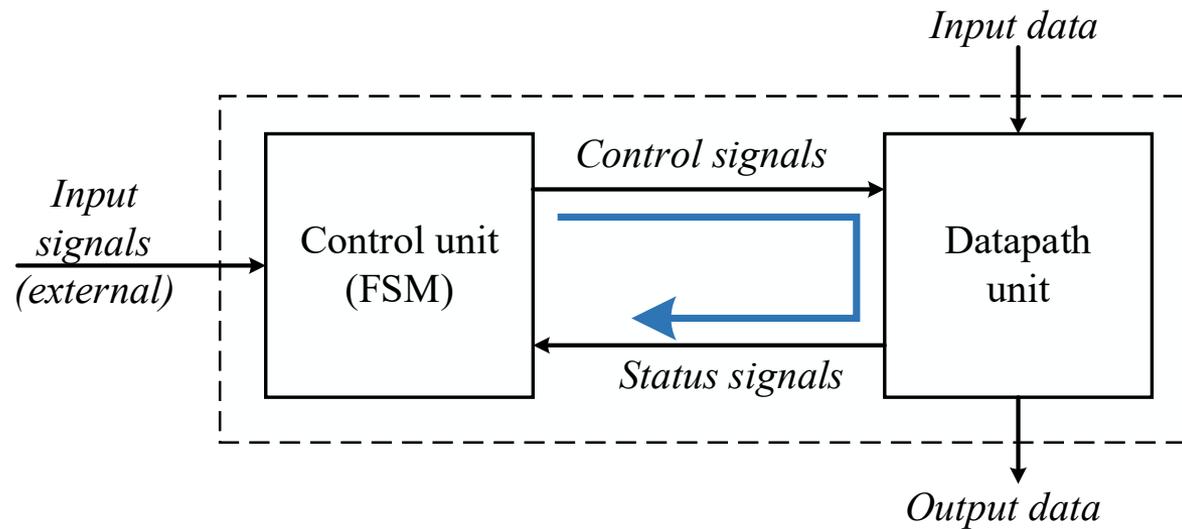
# ASM and Datapath (ASMD) Charts

---

- To form an ASMD: modify ASM (i.e. controller) by annotating each of its paths to indicate the concurrent register operations (i.e. datapath operations) when the controller makes a transition along the path
  - Not in conditional boxes
  - Not in state boxes
  - Because the datapath registers are not part of the controller
    - Fact: output generated by the controller controls the datapath register
- Clarify a design of a sequential machine by separating the design of its datapath from the design of the controller
- ASMD chart maintains a clear relationship between a datapath and its controller
  - Outputs generated by the controller control the datapath register
  - Outputs generated by datapath report the status of datapath back to the controller

# Control and Datapath

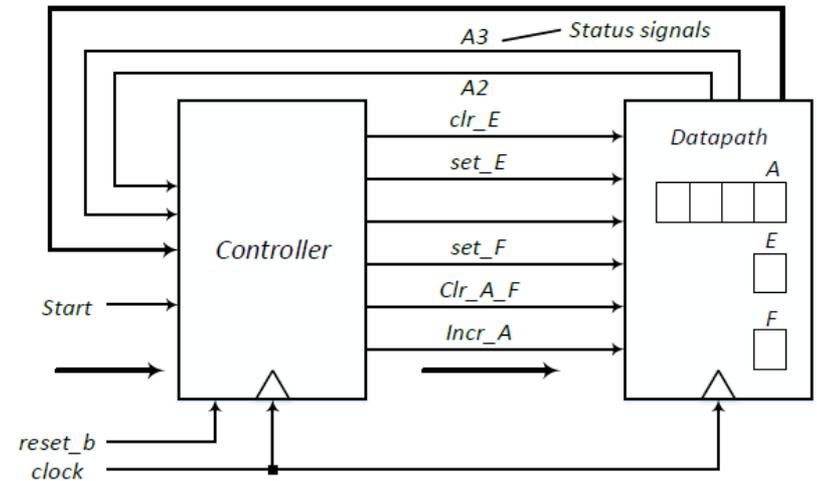
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- Most datapaths include arithmetic units. (e.g., adder, multiplier)
- The datapath unit manipulates data in registers according to the system's requirements.
- The control unit issues a sequence of commands to the datapath unit.
- The control logic be a finite state machine (FSM).

# 4-bit counter

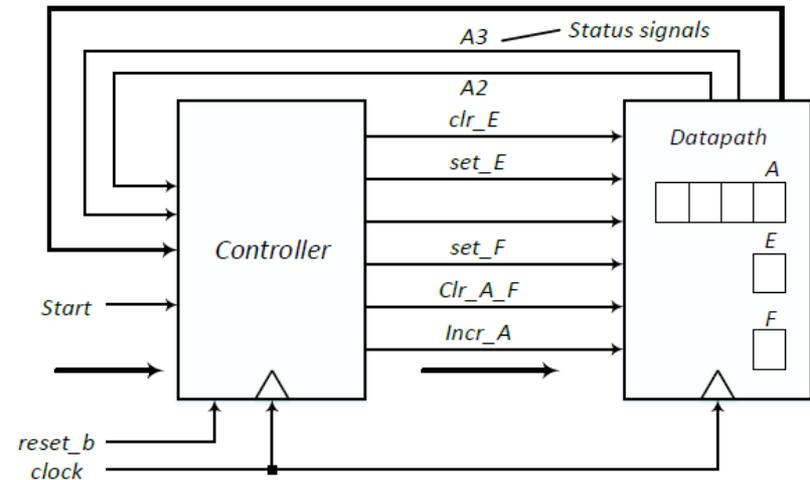
- A should reset to 0
  - Reset signal (reset\_b)
- System should stop counting at 2'b1101
- Control Signal to start and stop
- Uses flip-flops to store data
- Registers
  - A[3:0] – Contains count value
  - E – Control Value Flip-Flop
  - F – Finished State Flip-Flop
  - E, F, A[2], and A[3] are used to determine when the counter will stop counting



Block diagram of design example

# Controller Description

- ✓ *Start*  
input – Begin counter (take out of reset state)
- ✓ *reset\_b*  
input – Reset Counter
- ✓ *clr\_E*  
E is cleared to 0
- ✓ *set\_E*  
E is set to 1
- ✓ *set\_F*  
F is set to 1
- ✓ *Clr\_A\_F*  
F and A are set to 0
- ✓ *Incr\_A*  
Increment the counter (used to pause the system)



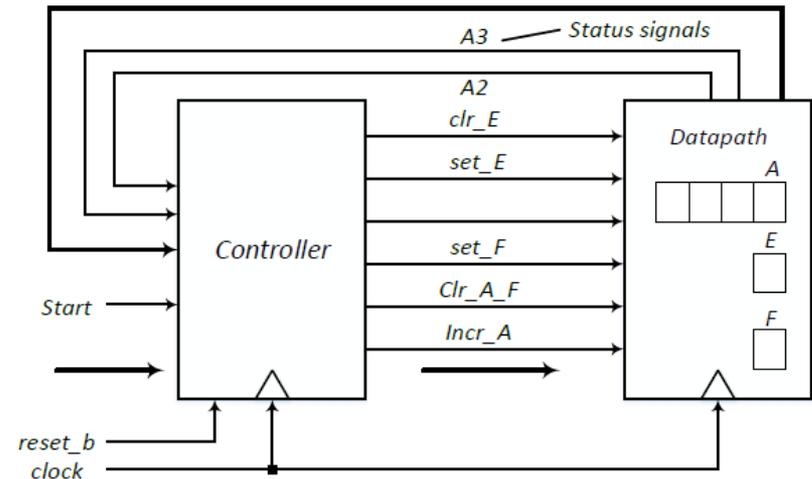
Block diagram of design example

# Datapath Signals

✓  $A[2]$

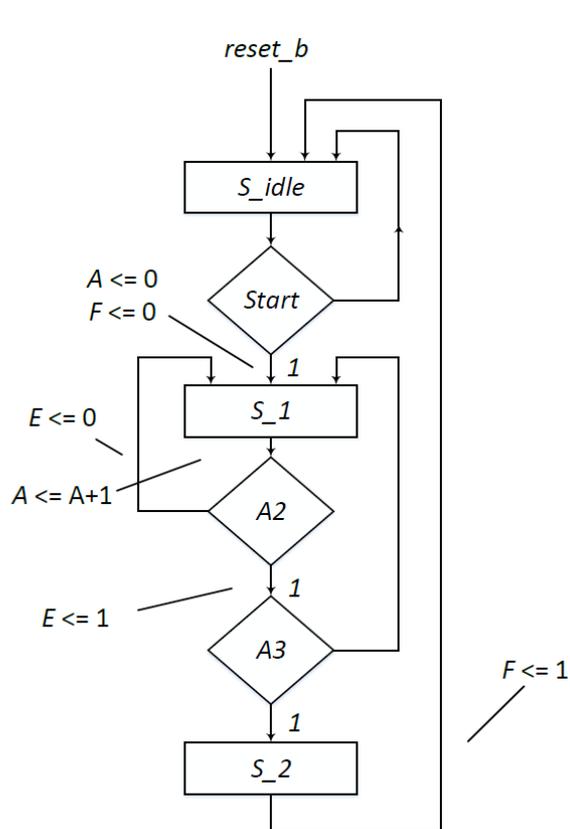
if  $A[2] = 0$  then E is assigned to 0 on the next clock pulse and system keeps counting

if  $A[2] = 1$  then E is assigned to 1 on the next clock pulse and  
If  $A[3] = 0$ , count continues  
If  $A[3] = 1$ , count stops, and F is assigned to 1

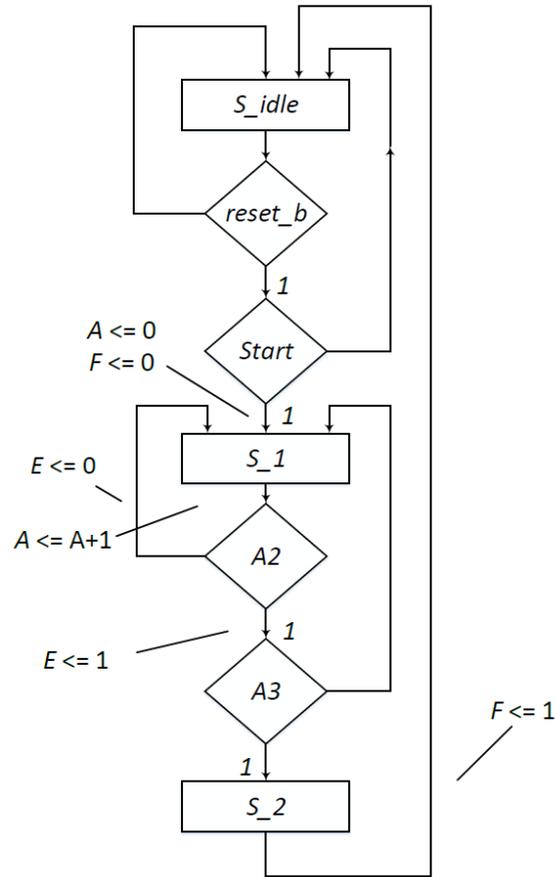


Block diagram of design example

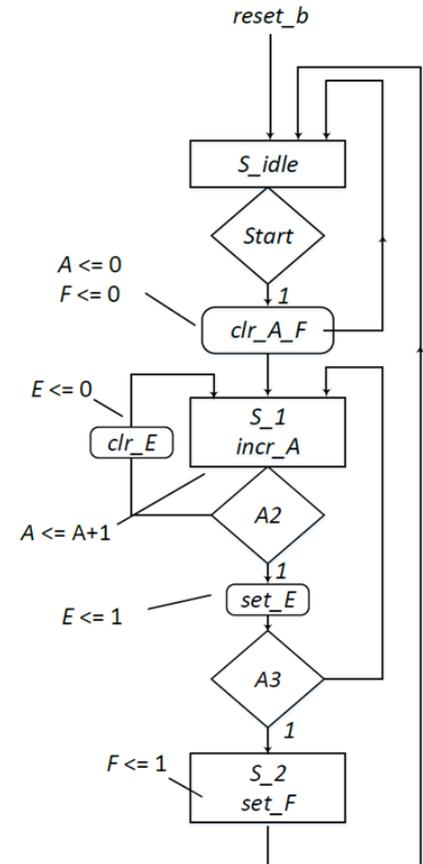
# ASM and ASMD Charts



ASM chart for controller state transitions, annotated with datapath register operations, **asynchronous reset**



ASM Chart for controller state transitions, annotated with datapath register operations, **synchronous reset**



ASMD chart for a completely specified controller, identifying datapath operations and associated control signals, **asynchronous reset**

# Verilog Code

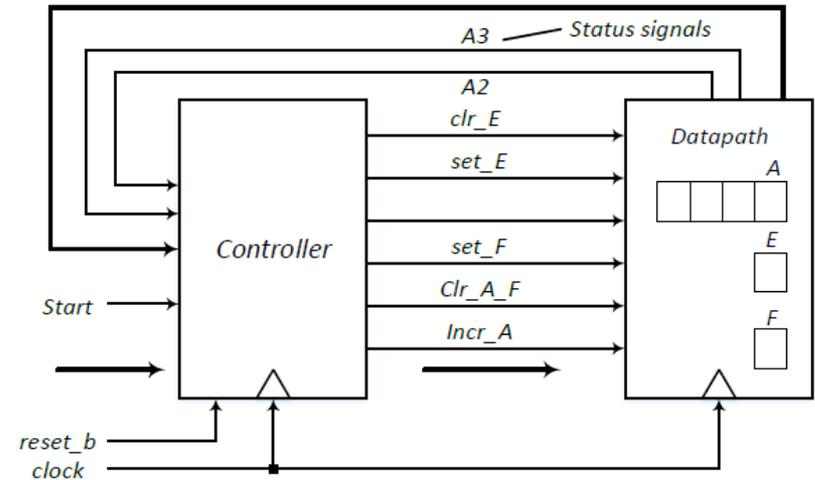
```
//RTL Description of design example
module Design_Example_RTL
(A,E,F,Start,clock, reset_b);
//Specify ports of the top-level module of
the design
```

```
Output [3:0] A;
Output E,F;
Input Start, clock, reset_b;
```

```
//instantiate controller and datapath units
```

```
Controller_RTL M0 (set_E, clr_E, set_F, clr_A_F, incr_A, A[2], A[3],
Start, clock, reset_b);
Datapath_RTL M1 (A, E, F, set_E, clr_E, set_F, clr_A_F, incr_A, clock);
```

```
endmodule
```



# Verilog: Controller

```
module Controller_RTL (set_E, clr_E, set_F, clr_A_F, incr_A, A2, A3, Start, clock,
reset_b);
```

```
output reg set_E, clr_E, set_F, clr_A_F, incr_A;
```

```
input Start, A2, A3, clock, reset_b;
```

```
reg [1:0] state, next_state;
```

```
parameter S_idle = 2'b00, S_1 = 2'b01, S_2 = 2'b11; //State Codes
```

```
// State transitions (edge sensitive)
```

```
always@ (posedge clock, negedge reset_b)
```

```
if (reset_b == 0) state <= S_idle;
```

```
else state <= next_state;
```

```
// Code next_state logic directly from ASMD chart
```

```
always @(state, Start, A2,A3) begin //Next_state logic (level sensitive)
```

```
next_state = S_idle;
```

```
case (state)
```

```
  S_idle: if (Start) next_state = S_1; else next_state = S_idle;
```

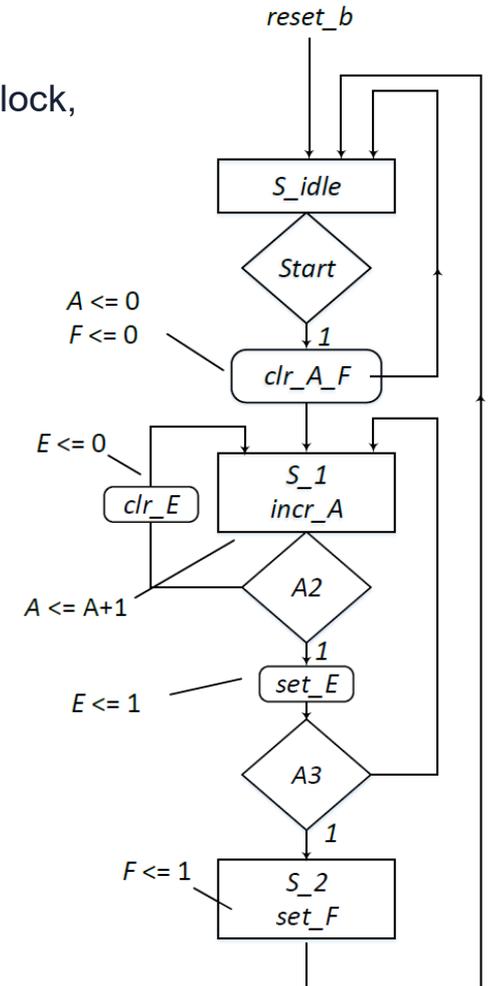
```
  S_1: if (A2& A3) next_state = S_2; else next_state = S_1;
```

```
  S_2:
```

```
    default: next_state = S_idle;
```

```
endcase
```

```
end
```

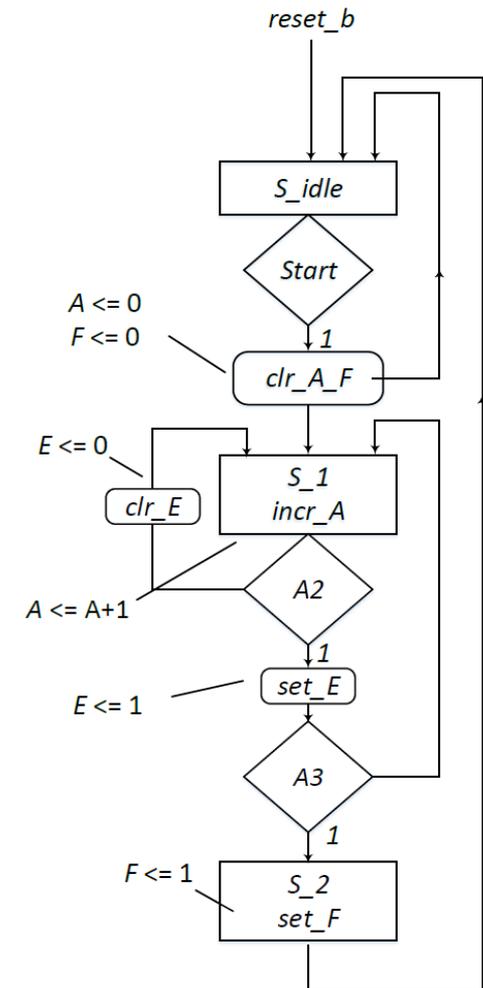


ASMD chart with asynchronous reset signal

# Verilog: Controller – Cont.

**//Code output logic directly from ASMD chart**

```
always @ (state, Start, A2) begin
    set_E    = 0; //Default assignments; assign by exception
    clr_E    = 0;
    set_F    = 0;
    clr_A_F  = 0;
    incr_A   = 0;
    case (state)
        S_idle: if (Start) clr_A_F = 1;
        S_1:   begin incr_A = 1; if (A2) set_E = 1; else clr_E = 1; end
        S_2:   set_F = 1;
    endcase
end
endmodule //End Controller Module
```



ASMD chart with asynchronous reset signal

# Verilog: Datapath

```

module Datapath_RTL (A, E, F, set_E, clr_E, set_F,
clr_A_F, incr_A, clock);
  output reg [3:0]    A; //Register for the counter
  output reg         E, F; //Flags
  input              set_E, clr_E, set_F, clr_A_F, incr_A, clock;

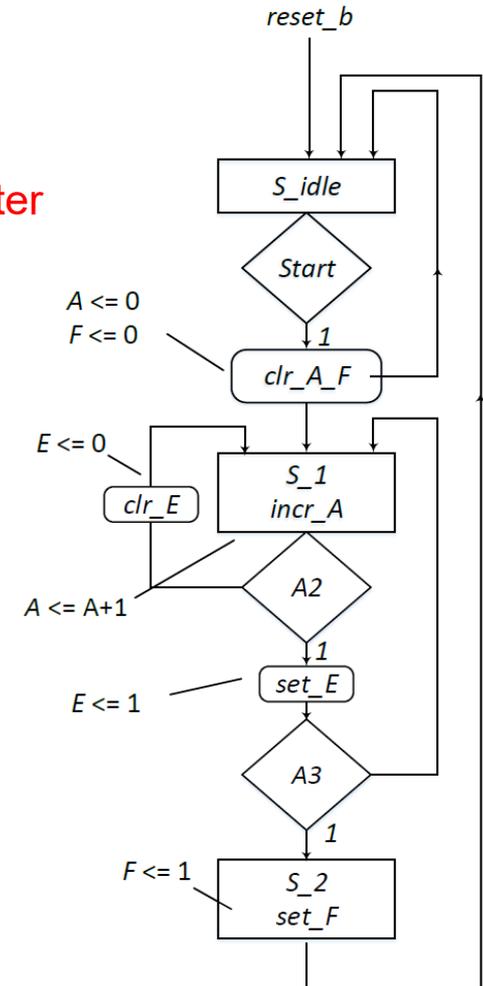
```

*// Code register transfer operations directly from ASMD chart.*

```

Always @ (posedge clock) begin
  if (set_E)      E <= 1;
  if (clr_E)      E <= 0;
  if (set_F)      F <= 1;
  if (clr_A_F)    begin A <= 0; F <= 0; end
  if (incr_A)    A <= A+1;
end
endmodule

```



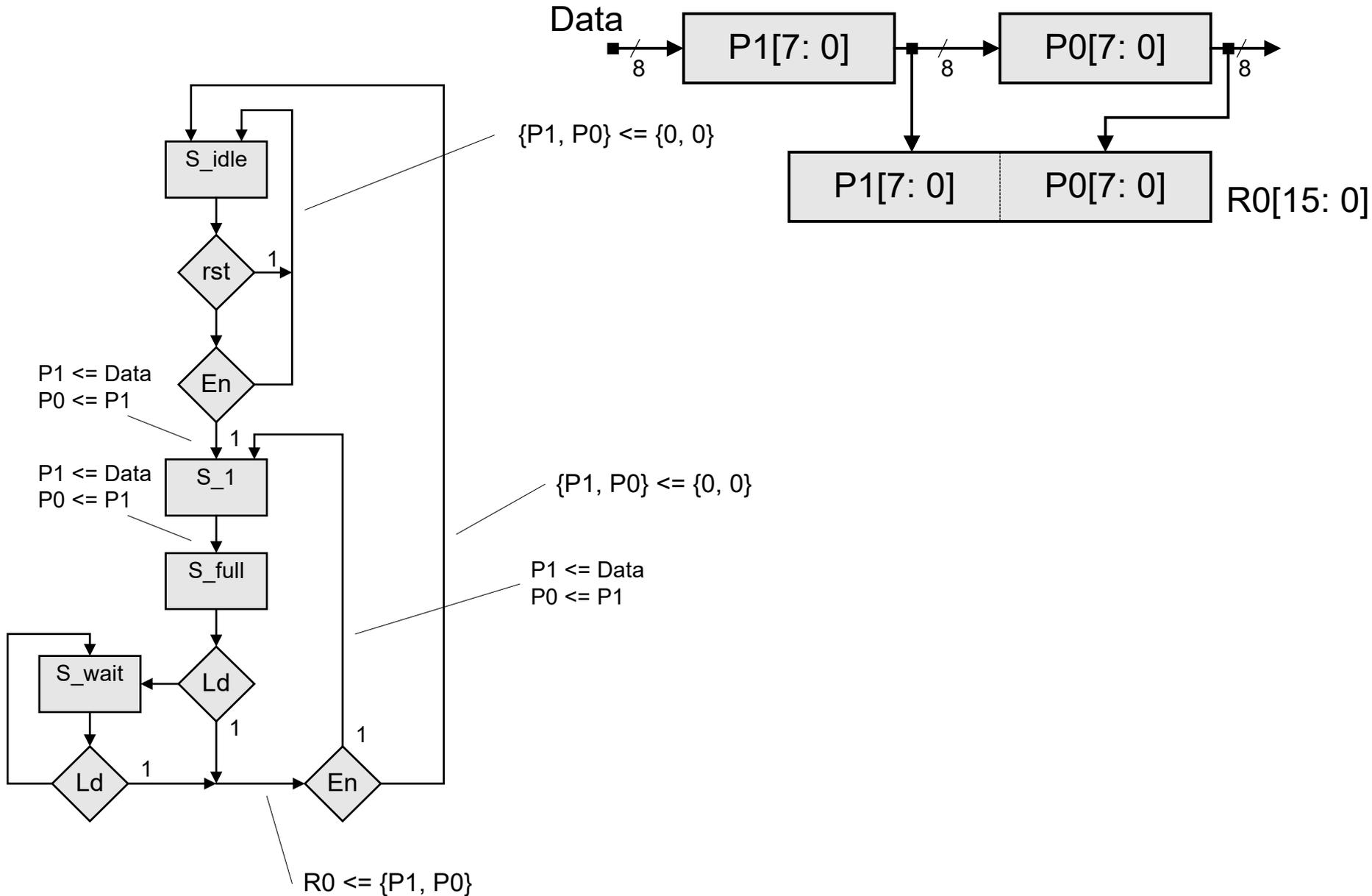
ASMD chart with asynchronous reset signal

# 2:1 Decimator Using 2-stage Pipeline

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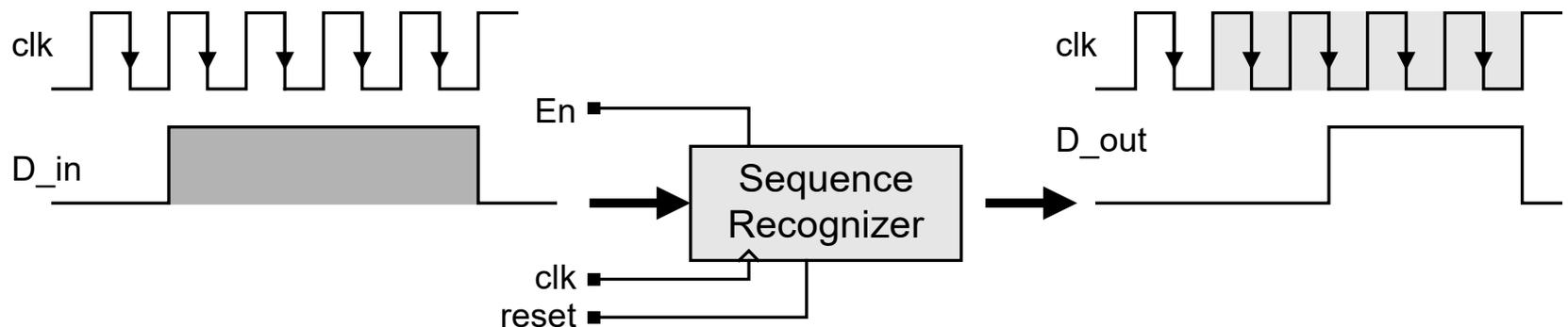
- Used to move data from a high clock rate datapath to a lower data rate datapath
  - Can also used to convert data from a parallel format to a serial format
- ASMD of the 2:1 decimator
  - A Mealy machine with synchronous reset to S\_idle
  - An incomplete ASMD
    - Because no conditional outputs
      - i.e., the output of the controller to control how datapath works
        - Such as adding an output for load-register
  - E.g. “Ld” state represents load to R0 since  $R0 \leftarrow \{P1, P0\}$  on the path leaving the state when Ld=1
  - Note that **datapath register operations made with a nonblocking assignment are concurrent**
    - Hence no race between  $R0 \leftarrow \{P1, P0\}$  and  $\{P1, P0\} \leftarrow \{0, 0\}$

# 2:1 Decimator Using 2-stage Pipeline (cont.)



# Synthesis of Sequence Recognizer

- Example: detect 3 consecutive 1s
  - Assert  $D\_out$  when a given pattern of consecutive bits has been received in its serial input stream,  $D\_in$
  - Apply data on the rising edge of the clock if the state transitions are to occur on the falling edge of the clock, and visa-versa
    - Recall the general rule for exercising FSM

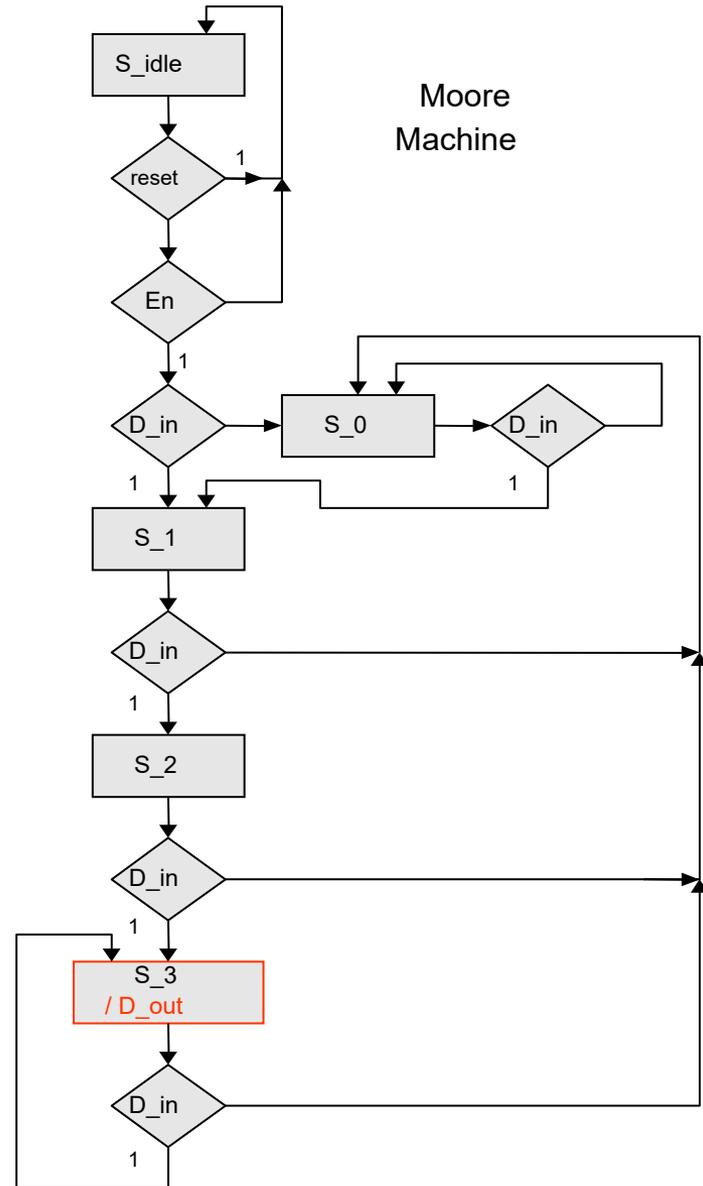
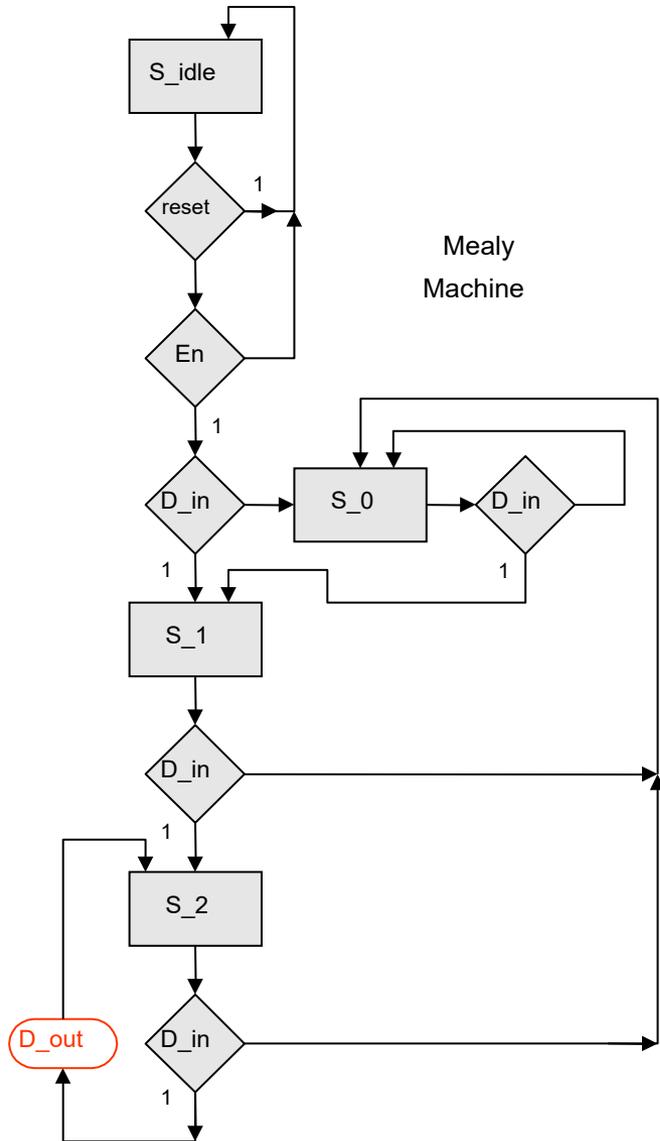


# Conventions to Describe Sequence Recognizers

---

- The output of a Mealy machine is valid immediately before the active edge of the clock controlling the machine
  - Data must be stable prior to active edge for at least the setup time
- Successive values inputs are received in successive clock cycles.
- A *non-resetting* machine continues to assert its output if the input bit pattern is overlapping
- A *resetting* machine asserts for one cycle after detecting the input sequence, and then de-asserts for one cycle before detecting the next sequence of bits

# Mealy and Moore ASMs (3 Consecutive 1s)



# Mealy and Moore for 3 Consecutive 1s (cont.)

---

- Both are non-resetting
  - How to modify them into resetting sequence recognizers?
- Moore has one more state than Mealy
- The Mealy machine anticipates  $D_{in}$  and asserts  $D_{out}$  before the third clock
- The Moore machine does not anticipate  $D_{in}$ 
  - That is, the Moore machine asserts  $D_{out}$  in the state reached after the third active edge of the clock

# Sequence Recognizer for 3 Consecutive 1s (cont.)

```

module Seq_Rec_3_1s_Mealy
    (D_out, D_in, En, clk, reset);

```

```

output      D_out;
input       D_in, En;
input       clk, reset;

```

```

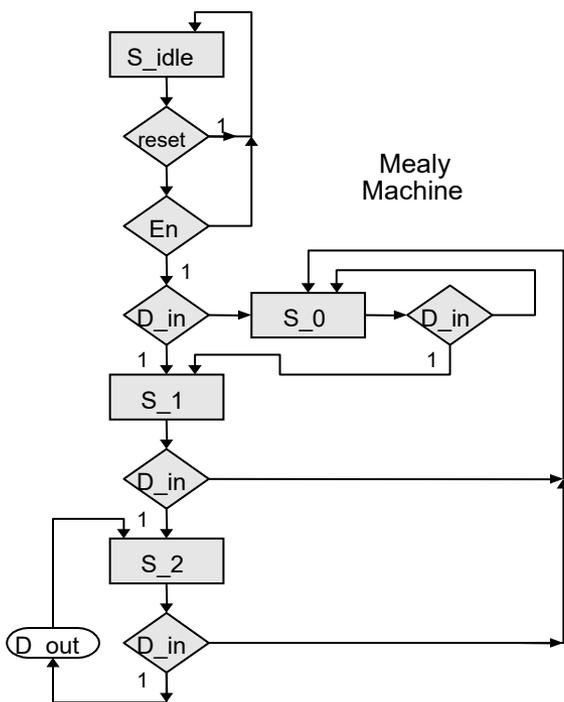
// Binary coding for states

```

```

parameter S_idle = 0;
parameter S_0 = 1;
parameter S_1 = 2;
parameter S_2 = 3;
reg[1: 0] state, next_state;

```



```

always @ (negedge clk)

```

```

    if (reset == 1) state <= S_idle; else state <= next_state;

```

```

always @ (state or D_in) begin

```

```

    case (state) // Partially decoded

```

```

        S_idle: if ((En == 1) && (D_in == 1)) next_state = S_1;

```

```

                else if ((En == 1) && (D_in == 0)) next_state = S_0;

```

```

                else next_state = S_idle;

```

```

        S_0: if (D_in == 0) next_state = S_0;

```

```

                else if (D_in == 1) next_state = S_1;

```

```

                else next_state = S_idle;

```

```

        S_1: if (D_in == 0) next_state = S_0;

```

```

                else if (D_in == 1) next_state = S_2;

```

```

                else next_state = S_idle;

```

```

        S_2: if (D_in == 0) next_state = S_0;

```

```

                else if (D_in == 1) next_state = S_2;

```

```

                else next_state = S_idle;

```

```

        default: next_state = S_idle;

```

```

    endcase

```

```

end

```

```

always @ (state or D_in) begin

```

```

    D_out = ((state == S_2) && (D_in == 1)); // Mealy output

```

```

end

```

```

endmodule

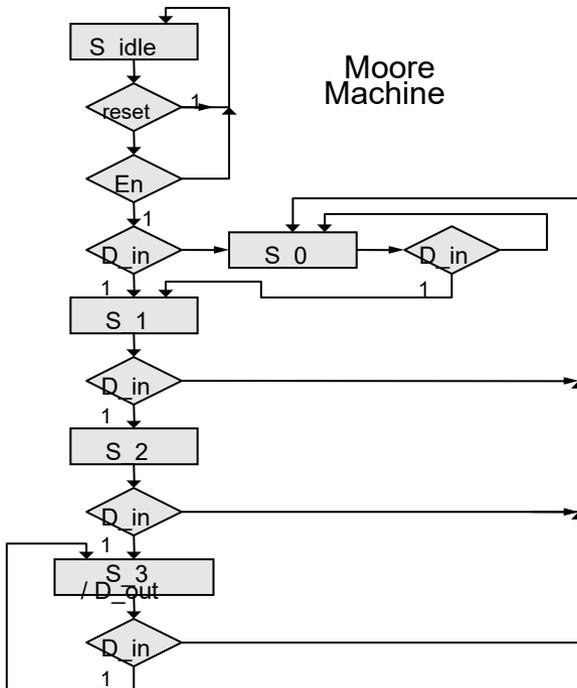
```

# Sequence Recognizer for 3 Consecutive 1s (cont.)

```

module Seq_Rec_3_1s_Moore
    (D_out, D_in, En, clk, reset);
output  D_out;
input   D_in, En;
input   clk, reset;
// Binary coding for states
parameter S_idle = 0;
parameter S_0 = 1;
parameter S_1 = 2;
parameter S_2 = 3;
parameter S_3 = 4;
reg[2:0] state, next_state;

```



```

always @ (negedge clk)
    if (reset == 1) state <= S_idle; else state <= next_state;

always @ (state or D_in) begin
    case (state)
        S_idle: if ((En == 1) && (D_in == 1))           next_state = S_1; else
                if ((En == 1) && (D_in == 0))           next_state = S_0;
                else                                     next_state = S_idle;
        S_0:     if (D_in == 0)                           next_state = S_0; else
                if (D_in == 1)                           next_state = S_1;
                else                                     next_state = S_idle;
        S_1:     if (D_in == 0)                           next_state = S_0; else
                if (D_in == 1)                           next_state = S_2;
                else                                     next_state = S_idle;
        S_2, S_3: if (D_in == 0)                           next_state = S_0; else
                if (D_in == 1)                           next_state = S_3;
                else                                     next_state = S_idle;
        default:
        endcase
    end

always @ (state) begin
    D_out = (state == S_3); // Moore output
end
endmodule

```

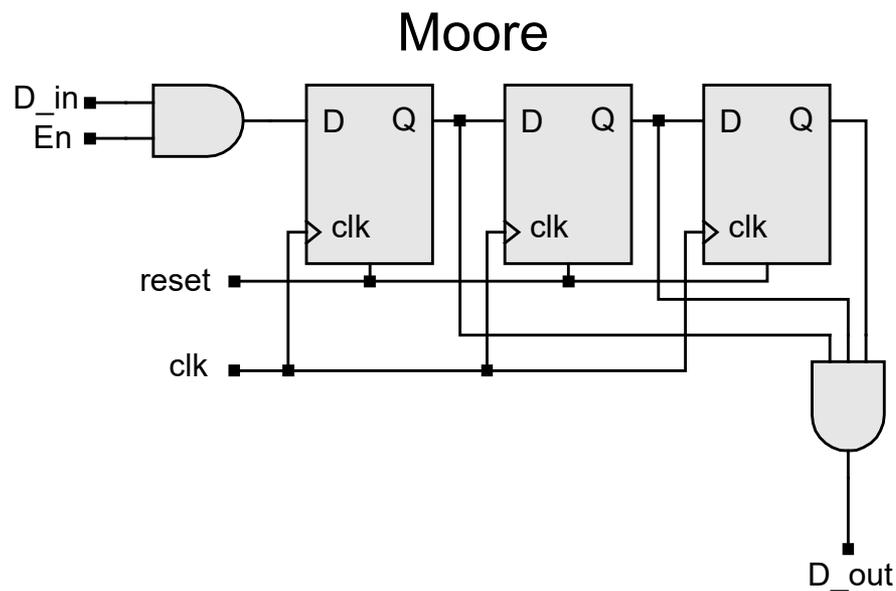
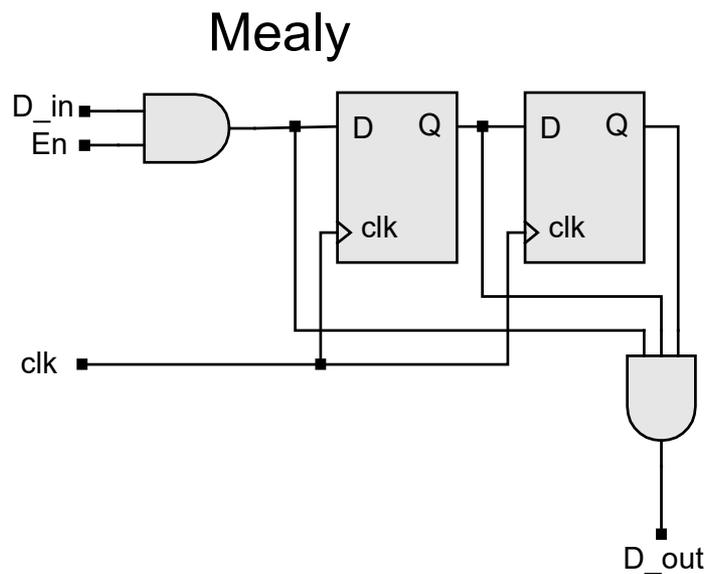
# Alternative Design for Sequence Recognizer

---

- Alternative approach: Shift input bits through a register and detect contents
  - Consider sequence recognizer as a datapath unit
  - Such as a shift register
  - Then compare the content of shift register with the expected pattern
- Note: an explicit state machine implementation of the alternative design for a sequence recognizer is not necessarily the most efficient implementation

# Alternative Design for Sequence Recognizer (cont.)

- The Mealy/Moore machines below are gated the datapath with  $En$ 
  - *What happens if  $En=0$ ?*
    - Register content will be lost
- Mealy has one less FF than Moore



# Alternative Design for Sequence Recognizer (cont.)

---

```
module Seq_Rec_3_1s_Mealy_Shft_Reg (D_out, D_in, En, clk, reset);  
  output          D_out;  
  input           D_in, En;  
  input           clk, reset;  
  parameter      Empty = 2'b00;  
  reg [1: 0]      Data;  
  always @ (negedge clk)  
    if (reset == 1) Data <= Empty; else if (En == 1) Data <= {D_in, Data[1]};  
  assign D_out = ((Data == 2'b11) && (D_in == 1 )); // Mealy output depends on primary input  
endmodule
```

```
module Seq_Rec_3_1s_Moore_Shft_Reg (D_out, D_in, En, clk, reset);  
  output          D_out;  
  input           D_in, En;  
  input           clk, reset;  
  parameter      Empty = 3'b000;  
  reg [2: 0]      Data;  
  always @ (negedge clk)  
    if (reset == 1) Data <= Empty; else if (En == 1) Data <= {D_in, Data[2:1]};  
  assign D_out = (Data == 3'b111); // Moore output depends on state only  
endmodule
```

# Design of a Datapath Controller

---

## 1. Understand the problem

- Especially the register operations that must execute on **a given datapath architecture**

## 2. Define ASM

- A state machine controlled by primary inputs and status of datapath register (i.e. the feedback linkage from datapath to controller)

## 3. Create ASMD

- Annotating ASM with datapath operations associated with state transitions (i.e. path) of the controller
- Register operation of ASMD written in register transfer notations with **NONBLOCKING assignments**
  - since they are executed concurrently in the datapath

## 4. Controller outputs to datapath

- **For Moore machines:** Annotate state of the controller with unconditional output signals (i.e. outputs of a state)
- **For Mealy machines:** Include conditional boxes for controller output signals to control datapath

## 5. Feedback linkage from datapath to controller

- If there are signals reports status of datapath back to the controller, then use decision box

## 6. Integration

- Integrate the verified datapath module and the verified controller module with one parent module to verify the overall functionality

# Counters and Registers

---

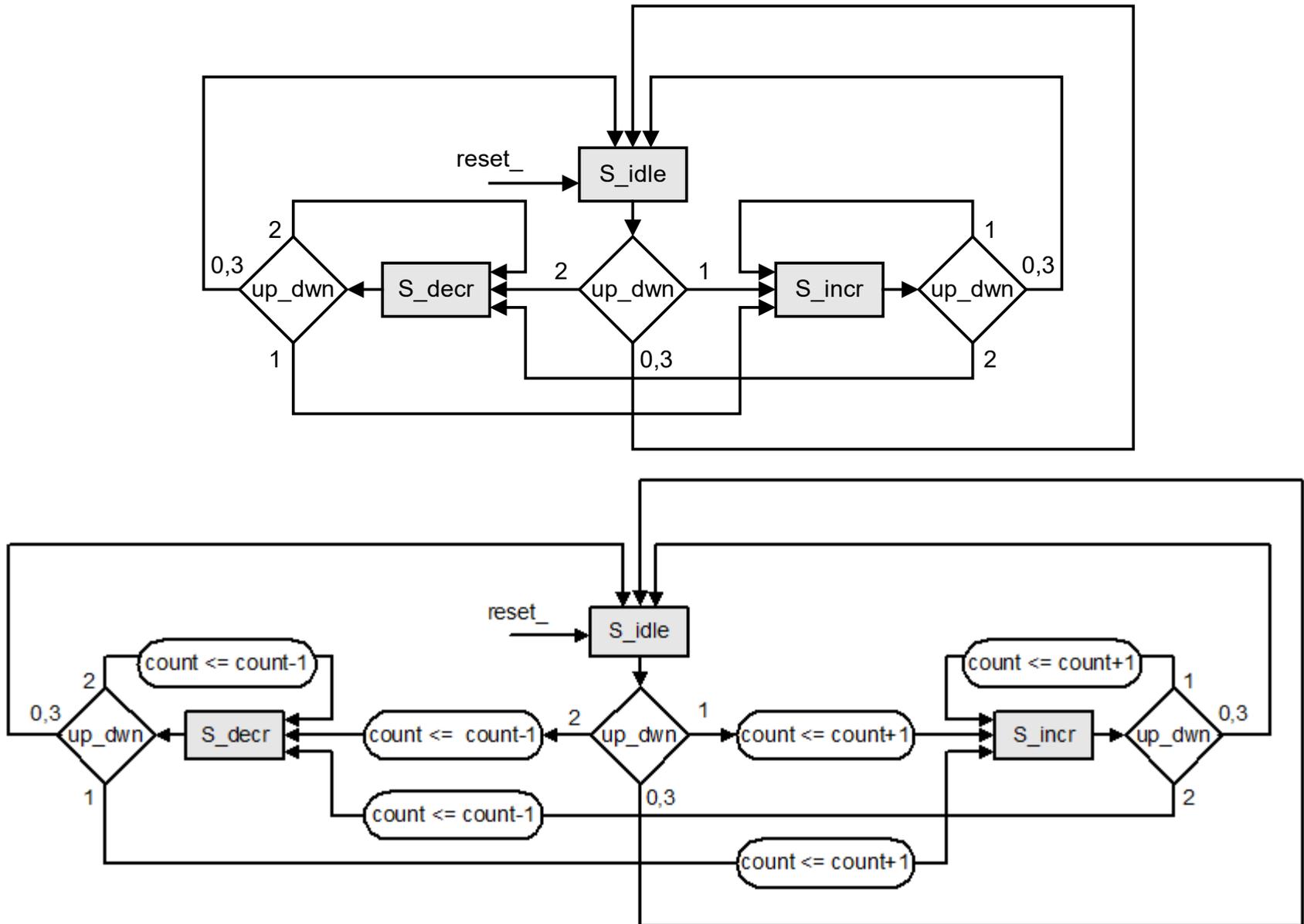
- Storage elements of counters and registers usually have the same synchronizing and control signals
  - One exception: ripple counter
    - Connects the output of a stage to the clock input of an adjacent stage
- Counters with asynchronous reset
- Ring counter
- Up/down/load counter
- Shift register
- Parallel load register
- Universal shift register
- Register file

# Counters

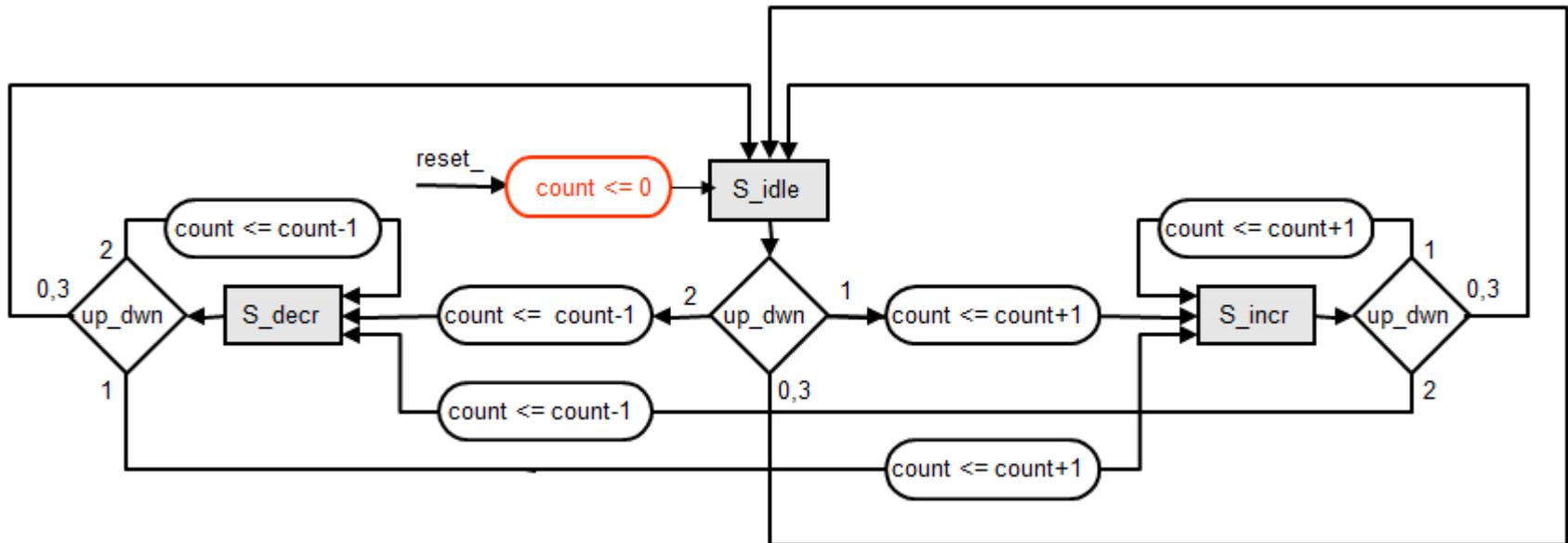
---

- The ASM/ASMD have no indication of the bit-width of the counter
- Three states: S\_idle, S\_incr and S\_decr
  - May be further simplified to a single state, S\_running
- 2-bit input up\_down to count up(1), count down(2) and hold the count (0 and 3)
- Active low asynchronous reset

# Counters(3 states) with Async Reset\_



# Counters (cont.)



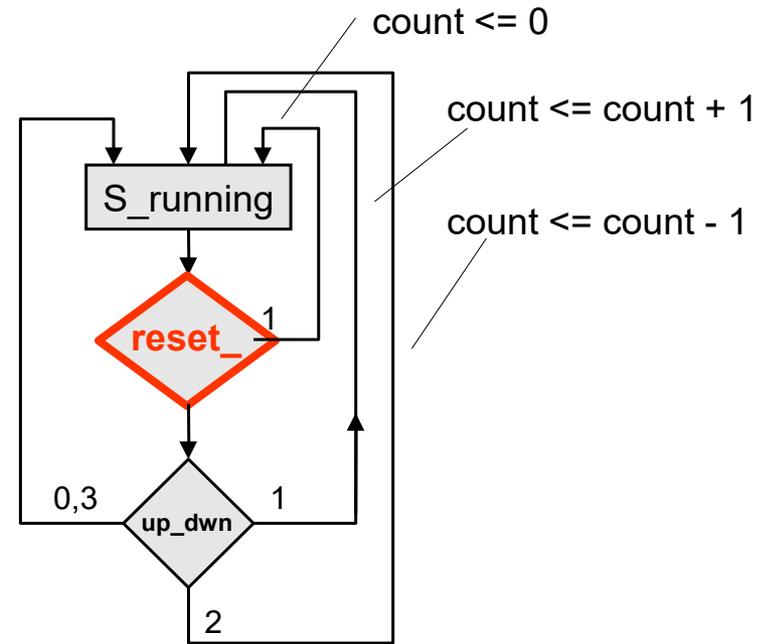
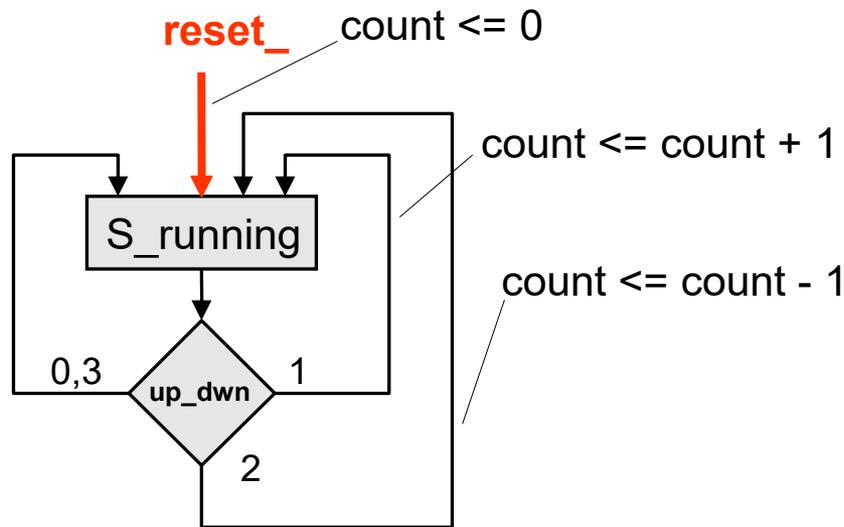
# Counters (3 states) with Async Reset\_ (cont.)

---

- It is an implicit state machine
  - No explicit states (S\_idle, S\_incr, S\_decr) used in the design
  - Implemented using if-then-else within edge-sensitive behavior

```
module Up_Down_Implicit1 (count, up_dwn, clock, reset_);  
  output [2: 0]    count;  
  input  [1: 0]    up_dwn;  
  input          clock, reset_;  
  reg [2: 0] count;  
  
  always @ (negedge clock or negedge reset_)  
    if (reset_ == 0)  
      count <= 3'b0;  
    else if (up_dwn == 2'b00 || up_dwn == 2'b11)  
      count <= count;  
    else if (up_dwn == 2'b01)  
      count <= count + 1;  
    else if (up_dwn == 2'b10)  
      count <= count - 1;  
  
endmodule
```

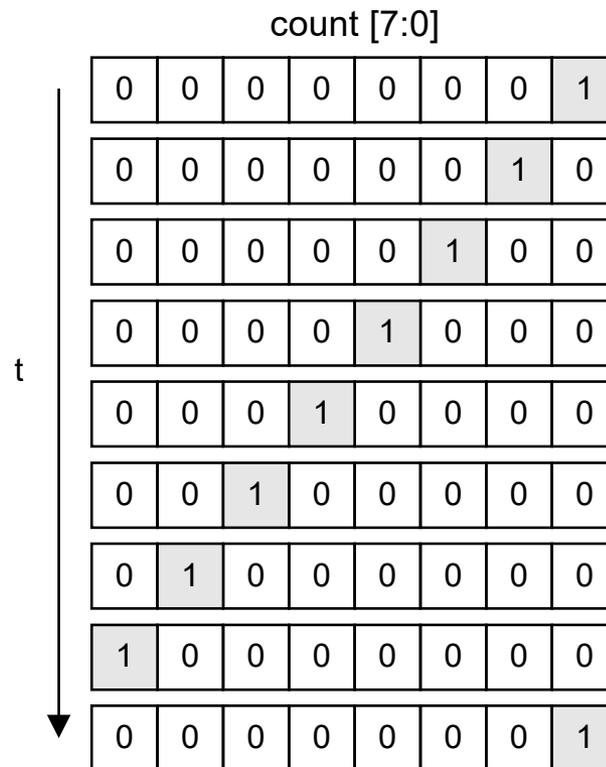
# Simplified Counter ASMDs with Async/Sync Reset\_



# Ring Counter

---

- Ring counter asserts a single bit that circulates through the counter in a synchronous manner



# Ring Counter (cont.)

---

- Activity of the machine is the same in every clock cycle
- This implementation is an implicit state machine

```
module ring_counter (count, enable, clock, reset);
  output          [7: 0]    count;
  input           enable, reset, clock;
  reg             [7: 0]    count;

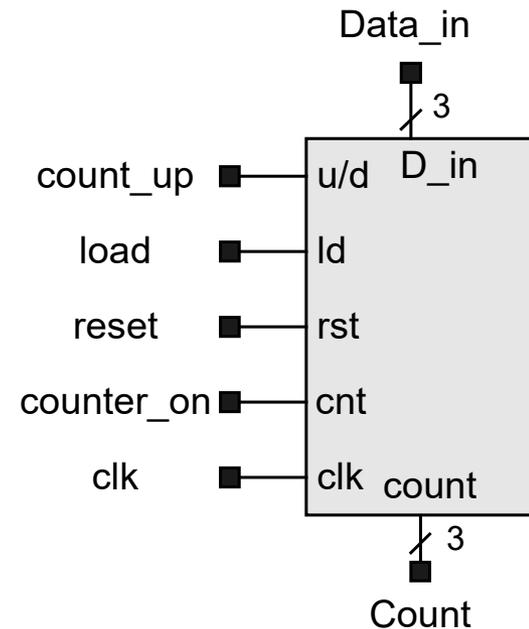
  always @ (posedge reset or posedge clock)
    if (reset == 1'b1) count <= 8'b0000_0001; else
      if (enable == 1'b1) count <= {count[6: 0], count[7]}; // Concatenation operator
endmodule
```

```
VHDL:: count <= count[6: 0] & count[7];
```

# Up/Down/Load Counter

---

```
module up_down_counter (Count, Data_in, load, count_up, counter_on, clk, reset);  
  output          [2: 0]    Count;  
  input           load, count_up, counter_on, clk, reset,;  
  input          [2: 0]    Data_in;  
  reg            [2: 0]    Count;  
  
  always @ (posedge reset or posedge clk)  
    if (reset == 1'b1) Count <= 3'b0; else  
      if (load == 1'b1) Count <= Data_in; else  
        if (counter_on == 1'b1) begin  
          if (count_up == 1'b1) Count <= Count +1;  
          else Count <= Count -1;  
        end  
  endmodule
```

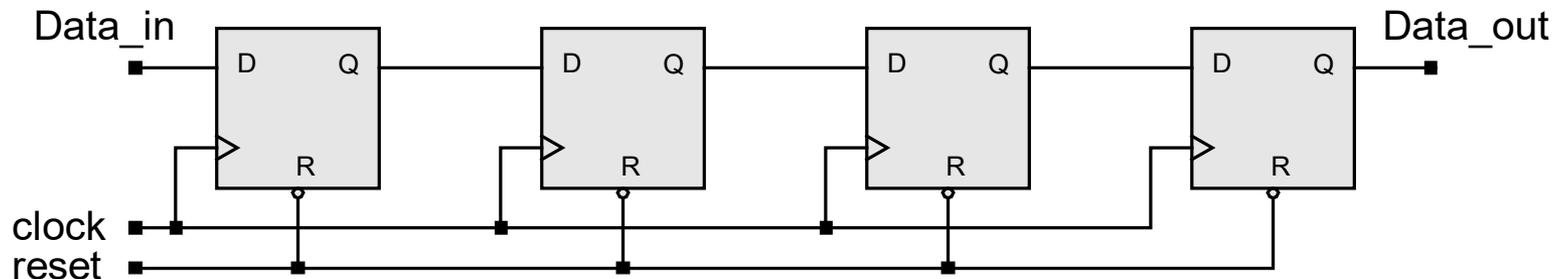


# Shift Register

---

- Remember the “model trap”
  - Must use **nonblocking** assignments in this design

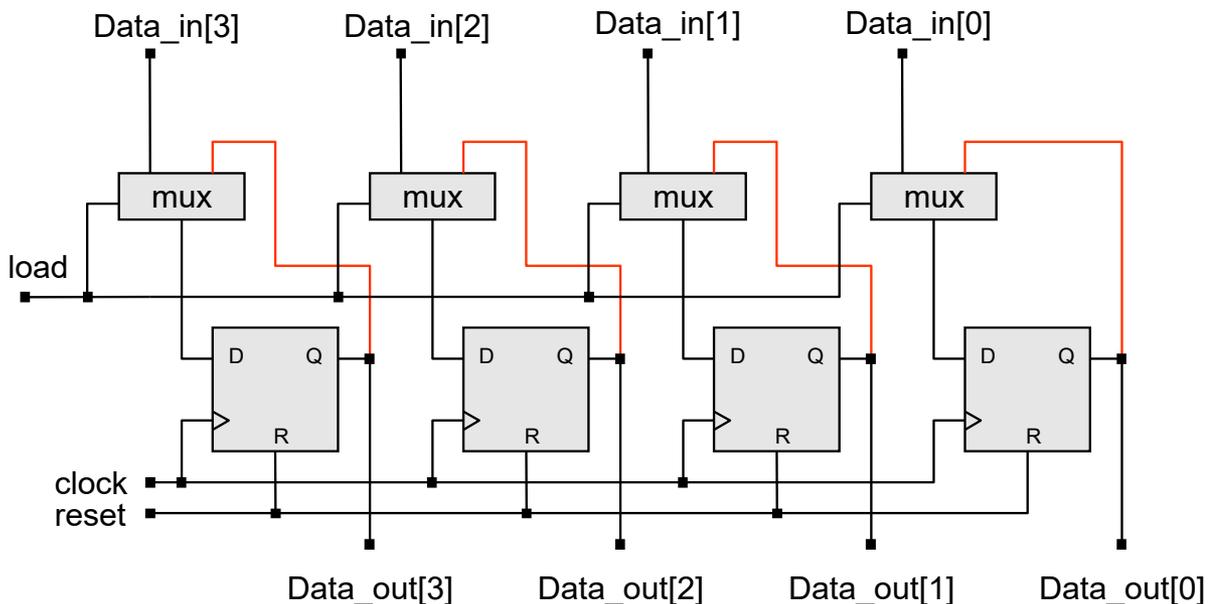
```
module Shift_reg4 (Data_out, Data_in, clock, reset);  
  output          Data_out;  
  input           Data_in, clock, reset;  
  reg [3: 0]      Data_reg;  
  
  assign Data_out = Data_reg[0];  
  
  always @ (negedge reset or posedge clock)  
  begin  
    if (reset == 1'b0) Data_reg <= 4'b0;  
    else                Data_reg <= {Data_in, Data_reg[3:1]}; //shift right  
  end  
endmodule
```



# Parallel Load Register

- MUX is synthesized from “else if (load==1'b1)”
  - How about “else” i.e. (load==1'b0)?
    - If not specified, retain the previous value

```
module Par_load_reg4 (Data_out, Data_in, load, clock, reset);
  input  [3: 0]  Data_in;
  input          load, clock, reset;
  output [3: 0]  Data_out;          // Port size
  reg          Data_out;          // Data type
  always @ (posedge reset or posedge clock)
  begin
    if (reset == 1'b1)
      Data_out <= 4'b0;
    else if (load == 1'b1)
      Data_out <= Data_in;
  end
endmodule
```



# Shift Registers

---

- Shift register with parallel load
  - later
- Arithmetic shift register
  - For signed number operation
    - MSB is preserved
  - Shift-left: multiply by 2
  - Shift-right: divide by 2

# Universal Shift Register

```
module Universal_Shift_Reg
```

```
(Data_Out, MSB_Out, LSB_Out, Data_In, MSB_In, LSB_In, s1, s0, clk, rst);
```

```
output [3: 0] Data_Out;
```

```
output MSB_Out, LSB_Out;
```

```
input [3: 0] Data_In;
```

```
input MSB_In, LSB_In;
```

```
input s1, s0, clk, rst;
```

```
reg Data_Out;
```

```
assign MSB_Out = Data_Out[3];
```

```
assign LSB_Out = Data_Out[0];
```

```
always @ (posedge clk) begin
```

```
if (rst) Data_Out <= 0;
```

```
else case ({s1, s0})
```

```
0: Data_Out <= Data_Out;
```

```
// Hold
```

```
1: Data_Out <= {MSB_In, Data_Out[3:1]};
```

```
// Serial shift from MSB
```

```
2: Data_Out <= {Data_Out[2: 0], LSB_In};
```

```
// Serial shift from LSB
```

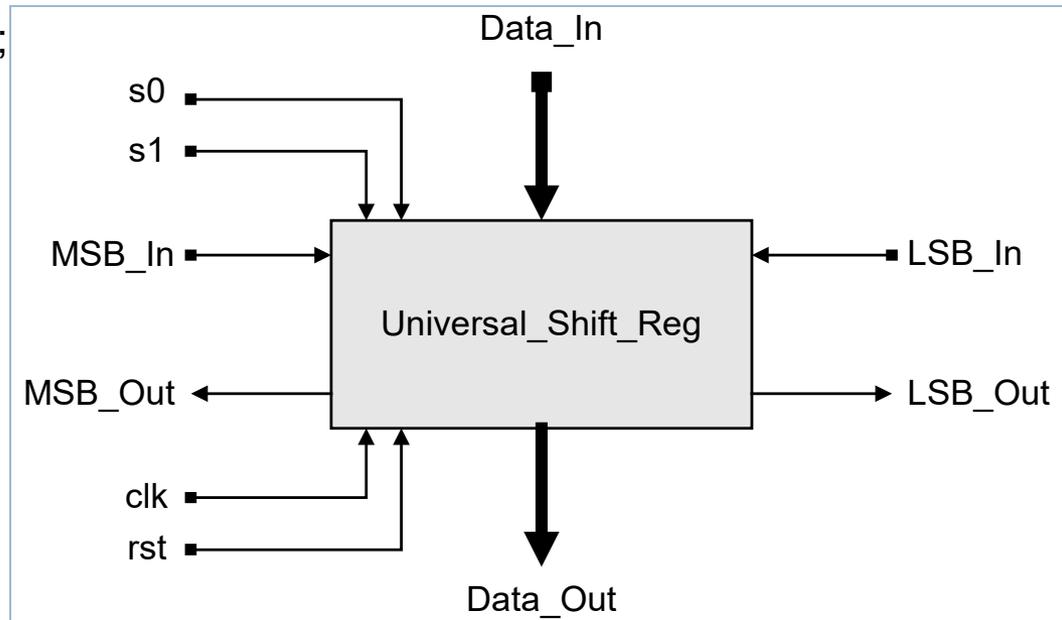
```
3: Data_Out <= Data_In;
```

```
// Parallel Load
```

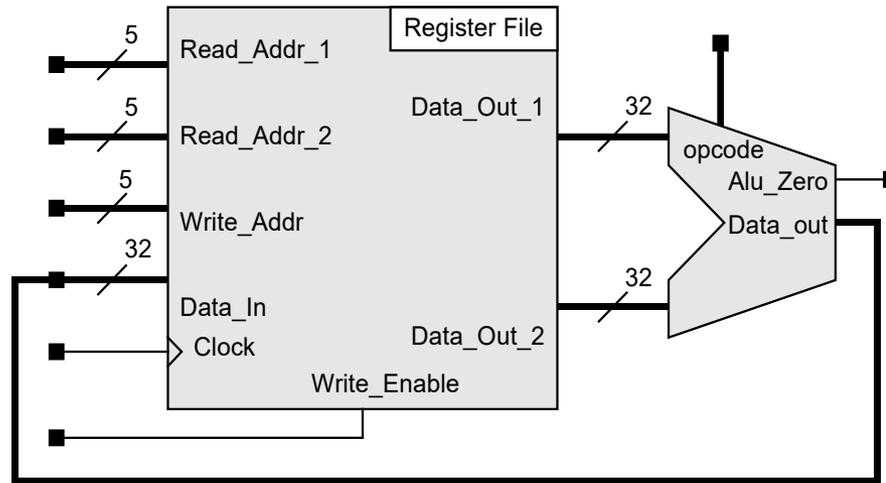
```
endcase
```

```
end
```

```
endmodule
```



# Register File



```

module Register_File (Data_Out_1,Data_Out_2,Data_in,
                      Read_Addr_1,Read_Addr_2,Write_Addr,Write_Enable,Clock);

```

```

output [31: 0]    Data_Out_1, Data_Out_2;

```

```

input  [31: 0]    Data_in;

```

```

input  [4: 0]     Read_Addr_1, Read_Addr_2, Write_Addr;

```

```

input  Write_Enable, Clock;

```

```

reg    [31: 0]    Reg_File [31: 0];    // 32bit x32 word memory declaration

```

```

assign Data_Out_1 = Reg_File[Read_Addr_1];

```

```

assign Data_Out_2 = Reg_File[Read_Addr_2];

```

```

always @ (posedge Clock) begin

```

```

    if (Write_Enable) Reg_File [Write_Addr] <= Data_in;

```

```

end

```

```

endmodule

```

```

type Reg is array (0 to 31) of std_logic_vector(31 downto 0);
signal Reg_File : Reg;

```

# “Concept of Memory” in Verilog

---

- Memory
  - Declaration an array of words
  - E.g. `reg [31:0] data_out; // one 32-bit word`  
`reg [31:0] Reg_file [31:0]; // 32x32 bit word memory`
- Verilog does not support 2-dimensional array
  - However, a word in a Verilog memory can be addressed directly
    - E.g., `Reg_file [12]`
  - A cell bit in a word can also be addressed indirectly by first loading the word into a buffer register then addressing the bit of the word
    - E.g. `Data_out = Reg_file [12];`  
`Data_out [1:0]`
- Decoder are synthesized automatically by synthesis tool in `Reg_file[]` to decode the address which locates a specific register